



Hi3518 Hardware Design  
**User Guide**

Issue            01  
Date             2013-06-30

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# About This Document

## Purpose

This document describes the design recommendations for the schematic diagrams, printed circuit board (PCB), and board heat dissipation of the Hi3518.

It also provides the hardware design methods of the Hi3518.

## Related Version

The following table lists the product version related to this document.

Product Name	Version
Hi3518	V100



## Intended Audience

This document is intended for:




- Technical support personnel
- Board hardware engineers

## Symbol Conventions

The symbols that may be found in this document are defined as follows.

Symbol	Description
 <b>DANGER</b>	Alerts you to a high risk hazard that could, if not avoided, result in serious injury or death.
 <b>WARNING</b>	Alerts you to a medium or low risk hazard that could, if not avoided, result in moderate or minor injury.



Symbol	Description
 <b>CAUTION</b>	Alerts you to a potentially hazardous situation that could, if not avoided, result in equipment damage, data loss, performance deterioration, or unanticipated results.
 <b>TIP</b>	Provides a tip that may help you solve a problem or save time.
 <b>NOTE</b>	Provides additional information to emphasize or supplement important points in the main text.

## Change History

Changes between document issues are cumulative. Therefore, the latest document issue contains all changes made in previous issues.

### Issue 01 (2013-06-30)

This issue is the first official release, which incorporates the following changes:

#### Chapter 1 Design Recommendations for Schematic Diagrams

In section 1.1.2, the descriptions related to internal reset are added.

In section 1.1.4, a note related to the NAND flash is added below Table 1-3.

Section 1.3.4 "RTC Interface" is added.

### Issue 00B06 (2013-02-28)

This issue is the sixth draft release, which incorporates the following changes:

#### Chapter 1 Design Recommendations for Schematic Diagrams

In section 1.1.5.2, the figures are updated, and the descriptions of the DDR parts of the Hi3518C\_BGA and Hi3518C\_QFP are added.

In section 1.3.3, a recommendation for excellent audio quality is added.

### Issue 00B05 (2013-02-05)

This issue is the fifth draft release, which incorporates the following changes:

#### Chapter 1 Design Recommendations for Schematic Diagrams

In section 1.1.5.5, descriptions on how to connect the two differential clock signals of the Hi3518C to a DDR3 SDRAM are added.

### Issue 00B04 (2012-12-26)

This issue is the fourth draft release, which incorporates the following changes:

#### Chapter 1 Design Recommendations for Schematic Diagrams

In section 1.2.2, the DDR power supply design is modified.



### **Issue 00B03 (2012-11-25)**

This issue is the third draft release, which incorporates the following changes:

#### **Chapter 2 PCB Design Recommendations**

In section 2.1.1, Figure 2-2 and related descriptions are added.

### **Issue 00B02 (2012-10-30)**

This issue is the second draft release, which incorporates the following changes:

#### **Chapter 1 Design Recommendations for Schematic Diagrams**

Figure 1-1 is updated.

In sections 1.1.3, 1.1.5.3, 1.2.1, 1.2.2, 1.3.2, 1.3.3, and 1.4, the descriptions are updated.

Section 1.2.5 "Power-on and Power-off Sequences" is added.

#### **Chapter 2 PCB Design Recommendations**

In sections 2.1.1, 2.1.3, 2.1.4, 2.1.5, 2.2.1, 2.2.2.1, and 2.2.2.2, the descriptions are updated.

The section "I/O Power" is added.

#### **Chapter 3 Heat Design Recommendations**

Table 3-1 and Table 3-2 are updated.

In section 3.2.1, the low-power configurations for idle modules in the schematic diagram are updated.

### **Issue 00B01 (2012-08-30)**

This issue is the first draft release.



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# 1 Design Recommendations for Schematic Diagrams

## 1.1 Design Recommendations for the Small System

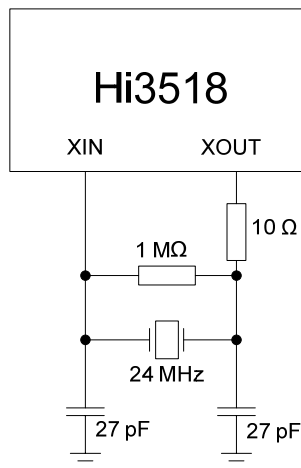
### 1.1.1 Clocking Circuit

System clocks can be generated in either of the following ways:

- Generate clocks by combining the internal feedback circuit of Hi3518 with an external 24 MHz crystal oscillator circuit.

Figure 1-1 shows the recommended connection mode of the crystal oscillator and other components.

Figure 1-1 Crystal oscillator circuit



- Generates clocks by using the external clock circuit and transmits clocks over the XIN pin.



## 1.1.2 Reset and Watchdog Circuit

The Hi3518 selects the internal or external reset mode by checking the status of the POR\_SEL pin during power-on:

 **NOTE**

The Hi3518A supports various reset modes, whereas the Hi3518C supports only internal reset.

- When the POR\_SEL level is low, internal reset is selected. After the master chip is powered on, the power-on-reset (POR) circuit resets the entire chip (the pulse width of the reset signal is about 140 ms), and the RSTN pin is invalid.

When internal reset is selected, the RSTN and WDG\_RSTN pins can be floated. In this case, peripherals (such as the flash memory) must be reset before the system is reset or peripherals and the system are reset at the same time. Otherwise, the system cannot boot. In addition, when internal reset is selected and the WDG\_RSTN pin is floated, the watchdog generates a reset signal in the chip to reset the system, and the SYS\_RSTN\_OUT pin outputs a reset signal to reset peripherals. You are advised to output the reset signal by using the SYS\_RSTN\_OUT pin of the Hi3518 to reset peripherals. The SYS\_RSTN\_OUT pin of the Hi3518A or Hi3518C\_BGA is located in V18. The Hi3518C\_QFP supports only internal reset, and its SYS\_RSTN\_OUT pin is pin 82.

- When the POR\_SEL level is high, external reset is selected. In this case, the RSTN pin is a reset signal input pin. The valid reset signal must have low-level pulse, and the pulse width must be greater than 12 input clock cycles for the XIN pin. Typically, the pulse width of the reset signal is 100–300 ms.

During board design, if external reset is selected, you are advised to use a dedicated reset chip to generate a reset signal to ensure system stability.

When an exception occurs in the system, the WDG\_RSTN pin can generate a low-level pulse. The pulse triggers an external reset chip to generate a reset signal. The WDG\_RSTN pin cannot directly connect to the RSTN pin.

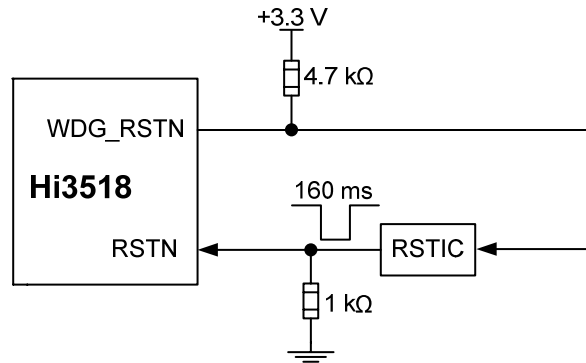


As the WDG\_RSTN pin is an open drain (OD) output pin, it must connect to an external pull-up resistor. A 4.7 k $\Omega$  resistor is recommended.

---

Figure 1-2 shows the typical external reset and watchdog circuit.

**Figure 1-2** Typical external reset and watchdog circuit



### 1.1.3 JTAG Debug Interface

The Hi3518 Joint Test Action Group (JTAG) interface complies with the IEEE1149.1 standard. The PC can connect to the Realview-ICE simulator over this interface for debugging the A9 CPU. [Table 1-1](#) describes the signals on the JTAG debug interface.

**Table 1-1** Signals on the JTAG debug interface

Signal	Description
TCK	JTAG clock input, internal pull-down. You are advised to connect a pull-down resistor to this signal on the board.
TDI	JTAG data input, internal pull-up. You are advised to connect a pull-up resistor to this signal on the board.
TMS	JTAG mode select input, internal pull-up. You are advised to connect a pull-up resistor to this signal on the board.
TRSTN	JTAG reset input, internal pull-down. When Hi3518 works properly, you are advised to connect a pull-down resistor to this signal on the board. When a debugger such as Realview-ICE is connected over the JTAG interface, you are advised to connect a pull-up resistor on the board.
TDO	JTAG data output. You are advised to connect a pull-up resistor to this signal on the board.

**NOTE**

For details about the impedance of pull-up and pull-down resistors described in [Table 1-1](#), see [Figure 1-3](#).

The Hi3518 can be set to normal mode or test mode by configuring the TEST\_MODE pin. For details, see [Table 1-2](#).

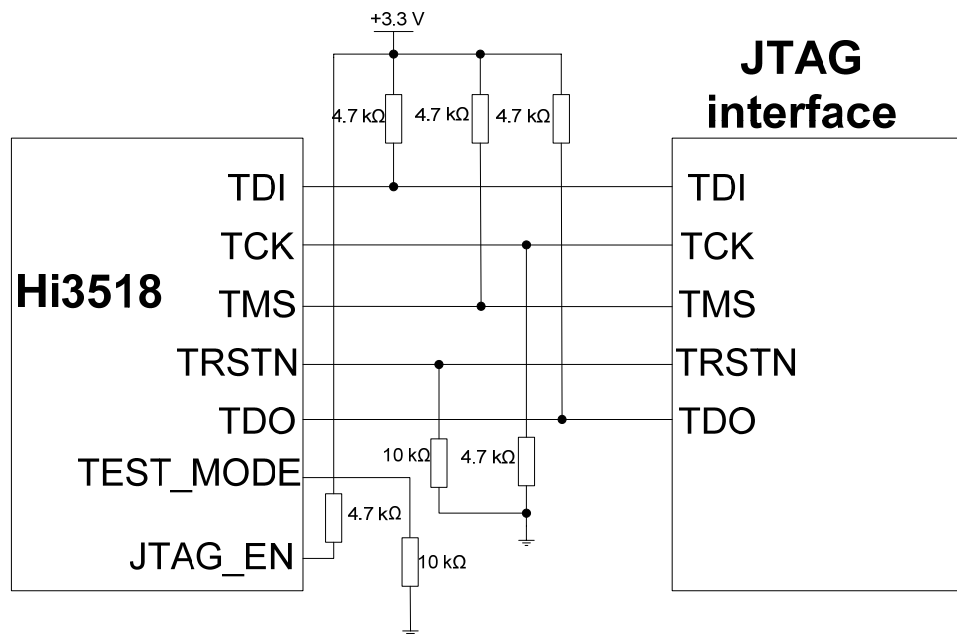


**Table 1-2** TEST\_MODE pin configuration

TEST_MODE	Description
0	The Hi3518 works in normal mode.
1	The Hi3518 works in test mode. In this case, the design for test (DFT) can be performed.

Figure 1-3 shows the JTAG connection mode and standard connector pins. To use the JTAG function, connect the JTAG\_EN pin on the board to a 4.7 kΩ pull-up resistor.

**Figure 1-3** JTAG connection mode and standard connector pins



### 1.1.4 System Configuration Circuit for Hardware Initialization

The Hi3518A can boot from the serial peripheral interface (SPI) flash or NAND flash and supports the NAND flash memories in various specifications. The Hi3518C can boot only from the SPI flash. The related pin must connect to pull-up or pull-down resistors as required during hardware initialization. Table 1-3 describes signal configurations.

**Table 1-3** Signal configurations

Signal	Direction	Description
JTAG_EN	I	JTAG debug enable. 0: disabled 1: enabled



Signal	Direction	Description
BOOT_SEL	I	Boot mode. 0: SPI flash 1: NAND flash Note: For the Hi3518C, BOOT_SEL must be set to 0 because the Hi3518C can boot only from the SPI flash.
NF_BOOT_PIN[4:0]	I	NAND type. 00001: 2 KB page size; 1-bit <a href="#">error correcting code</a> (ECC); 64-page block size 00011: 2 KB page size; 4-bit ECC; 64-page block size 00101: 2 KB page size; 24-bit ECC; 64-page block size 00110: 2 KB page size; 1-bit ECC; 64-page block size 01000: 4 KB page size; 4-bit ECC; 128-page block size 01001: 4 KB page size; 4-bit ECC; 64-page block size 01010: 2 KB page size; 4-bit ECC; 64-page block size 01011: 4 KB page size; 24-bit ECC; 128-page block size 01101: 8 KB page size; 24-bit ECC; 128-page block size 10000: 8 KB page size; 24-bit ECC; 64-page block size 10001: 4 KB page size; 24-bit ECC; 64-page block size 10011: 4 KB page size; 1-bit ECC; 64-page block size 10101: 2 KB page size; 4-bit ECC; 128-page block size 11001: 2 KB page size; 24-bit ECC; 128-page block size
SFC_ADDR_MODE	I	SPI flash controller (SFC) address mode. 0: 3-byte address mode 1: 4-byte address mode



**NOTE**

If the NANF flash is not used for the Hi3518A, NF\_BOOT\_PIN[0:3] must be floated. If the NAND flash is used for the Hi3518A, the value of NF\_BOOT\_PIN[4:0] must fall within the specified range.

## 1.1.5 DDR Circuits

### 1.1.5.1 Introduction

The double-data rate controller (DDRC) interface can be a DDR2 interface complying with the stub series terminated logic for 1.8 V (SSTL-18) standard or a DDR3 interface complying with the SSTL-15 standard.

The DDRC has the following features:



- Provides a DDRC interface and supports 8-bit or 16-bit data bus and 13-bit (for the Hi3518C) or 14-bit (for the Hi3518A) address bus. The Hi3518A DDRC supports a DDRn SDRAM chip select (CS), whereas the Hi3518C does not support CS.
- Supports 16-bit DDR2 SDRAM and 300–440 MHz DDR clock frequency. The maximum memory capacity is 256 MB (2 Gbits) for the Hi3518A or 128 MB (1 Gbit) for the Hi3518C.
- Supports 16-bit DDR3 SDRAM and 300–440 MHz DDR clock frequency. The maximum memory capacity is 256 MB (2 Gbits) for the Hi3518A or 128 MB (1 Gbit) for the Hi3518C.
- Supports various low-power modes for the DDRn SDRAM, such as power-down and self-refresh.

### 1.1.5.2 DDR Topology

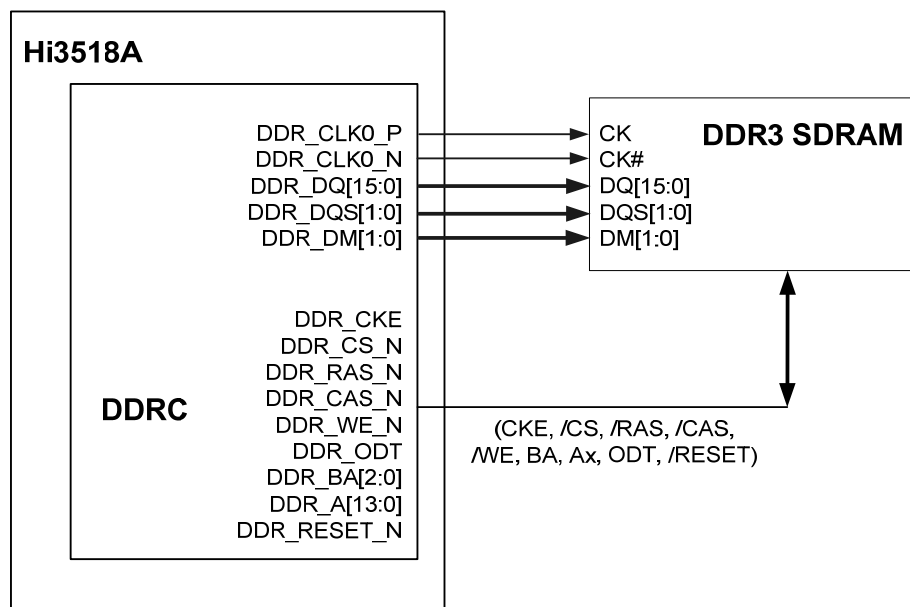


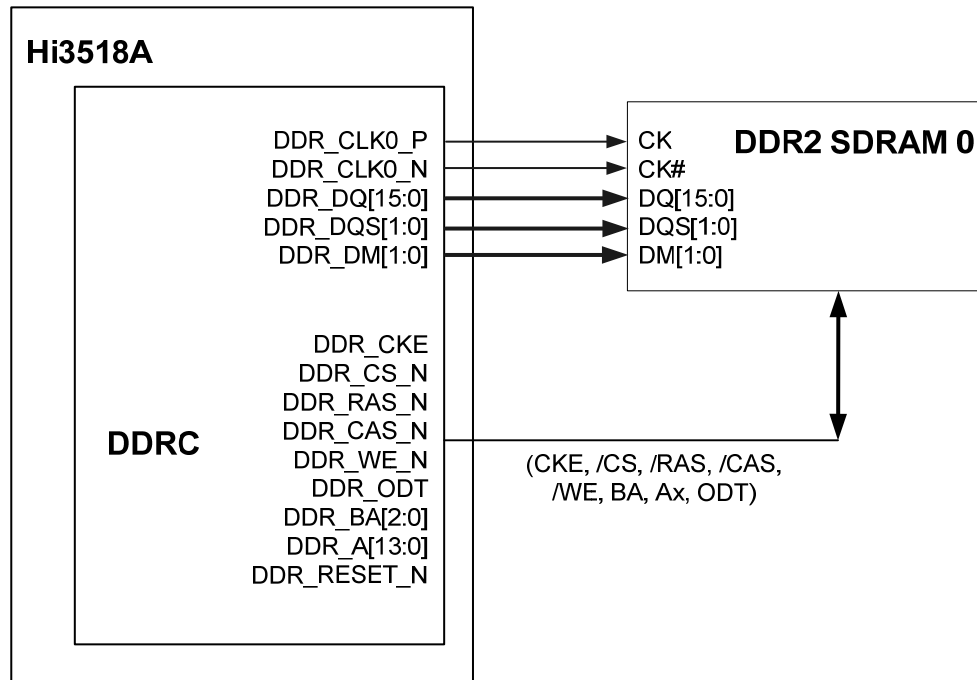
#### CAUTION

- The Hi3518C\_QFP does not have the DDR\_CS\_N pin. The DDR CS pin must connect to a 4.7 kΩ pull-down resistor. The DDR part of the Hi3518C\_QFP has only 13 address lines (A0–A12). The maximum capacity of the interconnected DDR is 1 Gbit.
- The DDR part of the Hi3518A has 14 address lines, whereas the DDR part of the Hi3518C\_BGA has only 13 address lines. The maximum capacity of the interconnected DDR is 1 Gbit.

Figure 1-4 shows the typical topology of connecting the Hi3518A to an external DDR3 SDRAM. Figure 1-5 shows the typical topology of connecting the Hi3518A to an external DDR2 SDRAM.

**Figure 1-4** Typical topology of connecting the Hi3518A to an external DDR3 SDRAM



**Figure 1-5** Typical topology of connecting the Hi3518A to an external DDR2 SDRAM

### 1.1.5.3 Design Recommendations for Matched Modes

#### Bidirectional DQ and DQS Signals

When the Hi3518 connects to a DDR, the DQ, DQS\_P, and DQS\_N signals of the Hi3518 connect directly to the DQ, DQS\_P, and DQS\_N signals of the DDR respectively. The details are as follows:

- When an external DDR2 SDRAM is connected:
  - If data is written to the DDR2 SDRAM (the Hi3518 acts as the output end), the DQ and DQS signals of the Hi3518 connect directly to the DQ and DQS signals of the DDR2 SDRAM respectively. The output impedance of the Hi3518 is 40  $\Omega$ , and the on-die termination (ODT) resistor in the DDR2 SDRAM is 75  $\Omega$ .
  - If data is read from the DDR2 SDRAM (the Hi3518 acts as the input end), the DQ and DQS signals of the Hi3518 connect directly to the DQ and DQS signals of the DDR2 SDRAM respectively. The output impedance of the DDR2 SDRAM is 40  $\Omega$ , and the ODT resistor in the Hi3518 is 75  $\Omega$ .
- When an external DDR3 SDRAM is connected:
  - If data is written to the DDR3 SDRAM (the Hi3518 acts as the output end), the DQ and DQS signals of the Hi3518 connect directly to the DQ and DQS signals of the DDR3 SDRAM respectively. The output impedance of the Hi3518 is 34  $\Omega$ , and the ODT resistor in the DDR3 SDRAM is 60  $\Omega$ .
  - If data is read from the DDR3 SDRAM (the Hi3518 acts as the input end), the DQ and DQS signals of the Hi3518 connect directly to the DQ and DQS signals of the DDR3 SDRAM respectively. The output impedance of the DDR3 SDRAM is 34  $\Omega$ , and the ODT resistor in the Hi3518 is 60  $\Omega$ .





## Differential Clocks

The following describes how to connect DDR\_CLK\_N and DDR\_CLK\_P signals to DDR2 or DDR3.

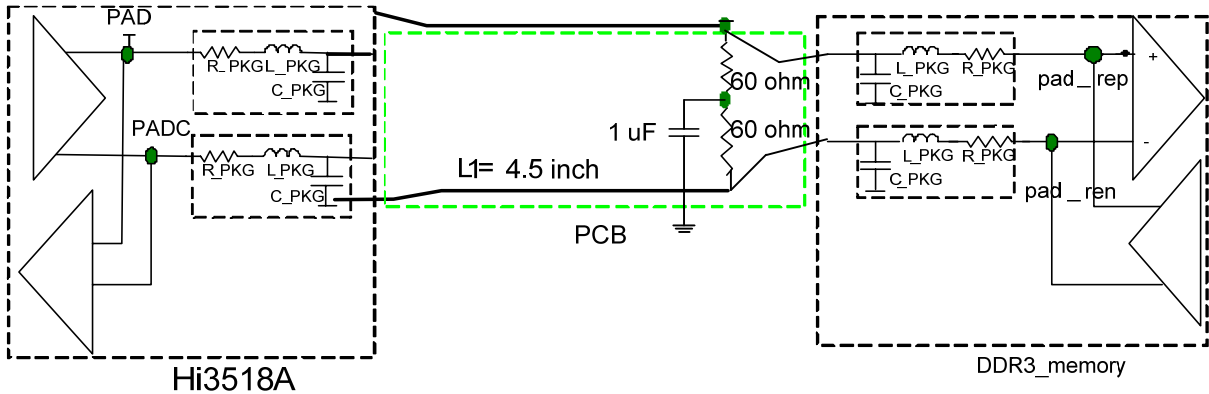
- Hi3518A

The mode of connecting clock signals to the DDR2 SDRAM is the same as that of connecting clock signals to the DDR3 SDRAM. In single-load mode, two 60 Ω±1% resistors are connected in series between DDR\_CLK\_N and DDR\_CLK\_P at the load end, as shown in Figure 1-6.

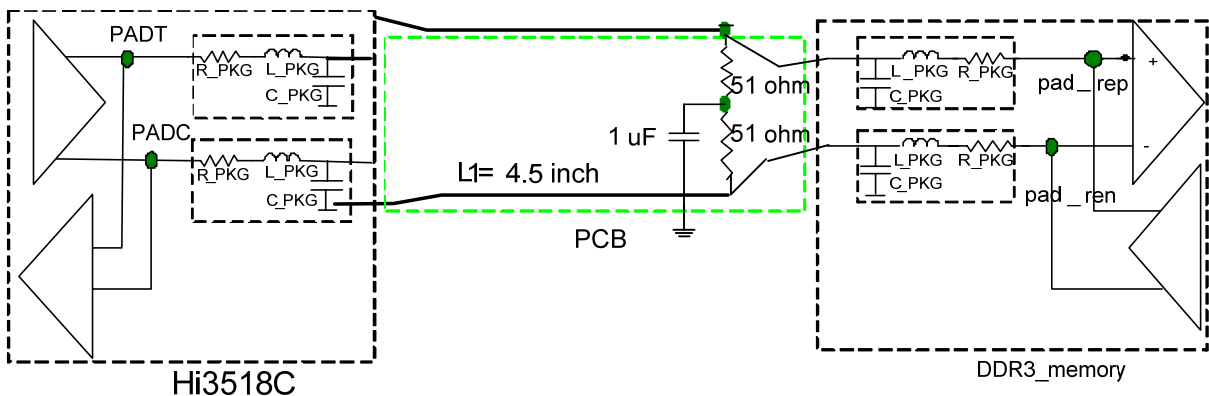
- Hi3518C

In single-load mode, two 51 Ω±1% resistors are connected in series between DDR\_CLK\_N and DDR\_CLK\_P at the load end, as shown in Figure 1-7.

**Figure 1-6** Hi3518A differential clocks DDR\_CLK\_N and DDR\_CLK\_P in one-drive-one mode when a DDR3 SDRAM is connected



**Figure 1-7** Hi3518C differential clocks DDR\_CLK\_N and DDR\_CLK\_P in one-drive-one mode when a DDR3 SDRAM is connected

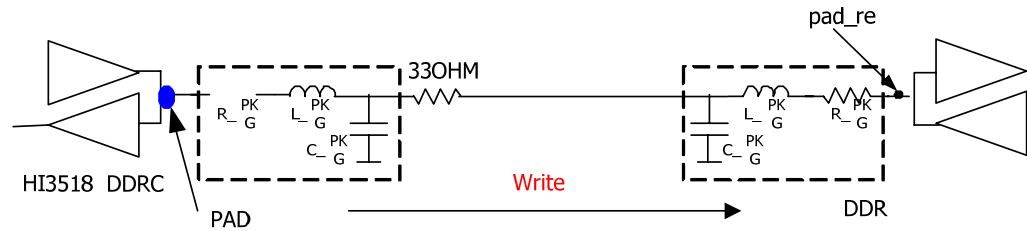


## Address Signals

The following takes a 16-bit DDR SDRAM as an example. The modes of connecting address signals are the same for the DDR2 and DDR3 SDRAMs.

A 33 Ω resistor is connected in series at the Hi3518 end, as shown in [Figure 1-8](#).

**Figure 1-8** Address signals in one-drive-one mode when a DDR2 or DDR3 SDRAM is connected



## Control Signals

The control signals of the DDR2 or DDR3 SDRAM connect directly to those of the Hi3518.

## DM Signals

The data mask (DM) signals of the DDR2 or DDR3 SDRAM connect directly to those of the Hi3518.

## 1.1.6 Design Recommendations for the Flash Schematic Diagrams

### 1.1.6.1 Introduction

The flash interface has the following features:

- When an SPI flash connects to the flash interface, the maximum capacity is MB SPI flash, and the clock frequency is 56 MHz.
- When a NAND flash connects to the flash interface, the maximum capacity is 512 GB.
- The multi-level cell (MLC) or single-level cell (SLC) NAND flash is supported, and the ECC mode is 1-, 4-, 8-, or 24-bit mode.
- The Hi3518 NAND flash controller (NFC) has two CSs for supporting two flash memories or a component with two CS pins. If the Hi3518A boots from the NAND flash, the NAND flash CS must connect to the NF\_CSN0 pin.
- The Hi3518C does not support the NAND flash.

### 1.1.6.2 Signal Processing

#### Matched Design for the SPI Flash

[Table 1-4](#) describes the recommended design when an external SPI flash is connected.

**Table 1-4** Recommended design when an external SPI flash is connected

Signal	4-Layer PCB	6-Layer PCB
SFC_CLK	A 33 Ω resistor is connected in series at the Hi3518 end.	A 33 Ω resistor is connected in series at the Hi3518 end.



Signal	4-Layer PCB	6-Layer PCB
SFC_DIO/SFC_DOI /SFC_WP/SFC_HOLD	The signal is connected directly, and SFC_WP must connect to a 4.7 k $\Omega$ pull-down resistor.	The signal is connected directly, and SFC_WP must connect to a 4.7 k $\Omega$ pull-down resistor.

## Matched Design for the NAND Flash

The NAND flash interface supports 8-bit SLC or MLC NAND flash.

Table 1-5 describes the recommended design when an external NAND flash is connected.

**Table 1-5** Recommended design when an external NAND flash is connected

Signal	4-Layer PCB	6-Layer PCB
NF_WEN/NF_REN	A 33 $\Omega$ resistor is connected in series at the Hi3518 end.	A 33 $\Omega$ resistor is connected in series at the Hi3518 end.
DQ[0:7] /NF_ALE/NF_CLE	The signal is connected directly.	The signal is connected directly.

## 1.2 Design Recommendations for Power Supplies

For details about system power supply design, see the schematic diagram of the Hi3518A demo board.

### 1.2.1 Core Power Supply

The core power pins DVDD12 connect to the 1.2 V digital power. The current of the power chip for DVDD12 must be greater than or equal to 1 A. Each DVDD12 pin connects to at least two 10  $\mu$ F filtering bypass capacitors and then to GND. At least one 100 nF decoupling capacitor connects to each DVDD12 pin close to the power pin.

### 1.2.2 DDR Power Supply

The DDRC and interface comply with the DDR3 SSTL-15 and SSTL-18 standards. The 1.5 V or 1.8 V power and 0.75 V or 0.9 V reference voltage Vref are required. The Hi3518 and DDR SDRAM must share the 1.5 V or 1.8 V power.

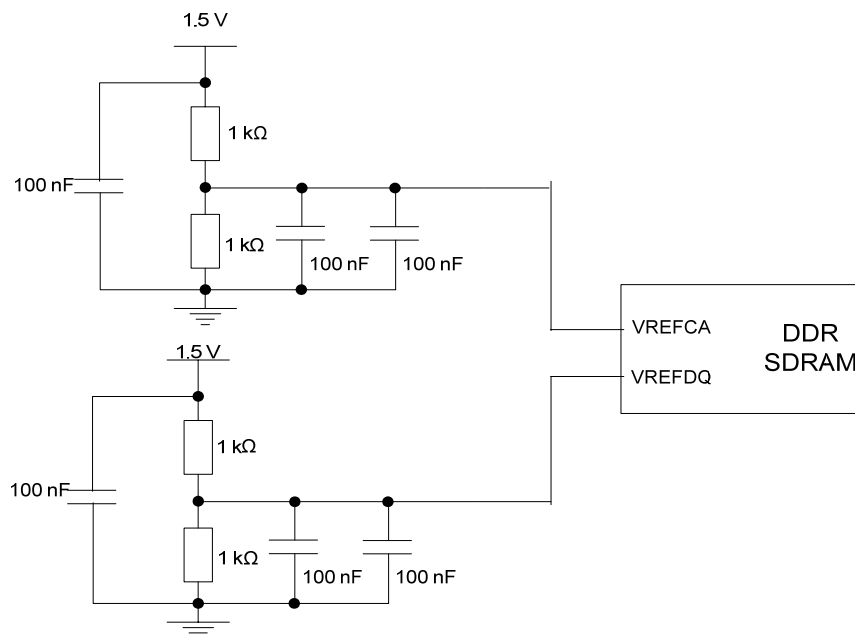
The DDR power pins DDR\_VDDQ connect to the 1.5 V or 1.8 V digital power. It is recommended that the Hi3518 DDR share the same power supply as that of the connected DDR.

A 100 nF ceramic filtering capacitor needs to be connected close to each power pin. At least one 10  $\mu$ F filtering capacitor connects to the power supply of the DDR3 SDRAM and then to GND.

You are advised to design a separate direct current (DC) to DC circuit on the board to supply power to the DDR2 SDRAM or DDR3 SDRAM and the 1.5 V or 1.8 V power pins of the DDRC. The power is supplied to the DDR3 SDRAM (0.75 V) or DDR2 SDRAM (0.9 V) and reference power pin Vref of the DDRC by using 1% 1 k $\Omega$  voltage-divider resistors. One 0.1  $\mu$ F decoupling capacitor is connected close to each power pin and reference power pin.

Figure 1-9 shows the reference design of the DDR3 voltage-divider circuit. The DDR2 voltage-divider circuit is similar to the circuit shown in Figure 1-14. The difference is that the power voltage needs to be changed to 1.8 V. **Note that power must be separately supplied to VREFCA and VREFDQ.**

Figure 1-9 Reference design of the DDR3 voltage-divider circuit



### 1.2.3 I/O Power Supply

The I/O power pins DVDD33 connect to the 3.3 V digital power. Each VDD33 pin connects to at least one 10  $\mu$ F filtering bypass capacitor and then to GND. In addition, a 100 nF decoupling capacitor connects to each DVDD33 pin close to the power pin.

For the video input (VI) interface power (DVDD3318 pin), the I/O power supplies for the sensor CLK, VI, and SPI0 parts can be 1.8 V or 3.3 V, and the I/O level standards of various sensors are supported.

### 1.2.4 PLL Power Supply

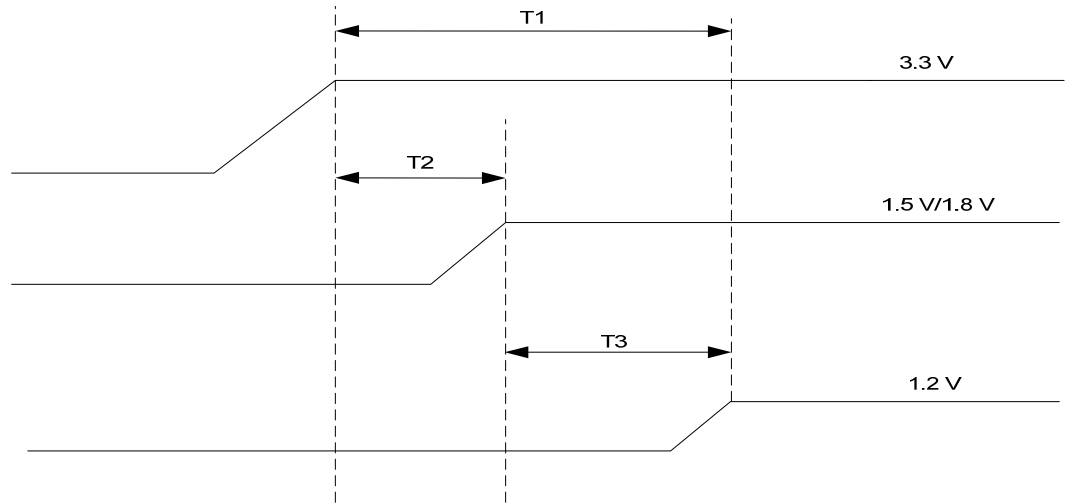
The phase-locked loop (PLL) power supply is isolated by using an electromagnetic interference (EMI) bead. For details, see the schematic diagram of the Hi3518A demo board.

### 1.2.5 Power-on and Power-off Sequences

There are requirements on the power-on and power-off sequences for the core power, DDR power, and I/O power. For details, see Figure 1-10, and Figure 1-11.



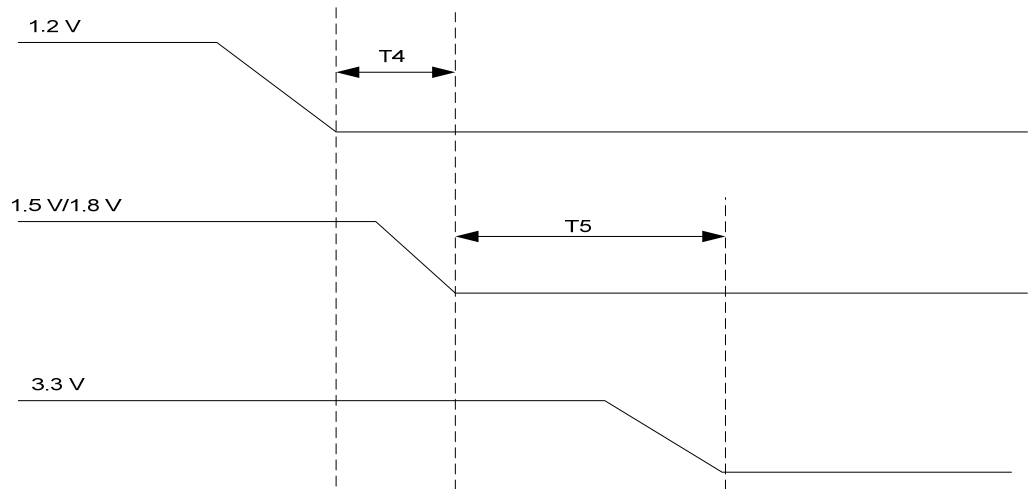
**Figure 1-10** Power-on sequence



**NOTE**

$0 < T1 \leq 100 \text{ ms}$ ,  $T2 > 0$ ,  $T3 > 0$

**Figure 1-11** Power-off sequence



**NOTE**

$T4 > 0$ ,  $T5 > 0$

## 1.2.6 Notes

Ensure that the output voltage of power supplies meets the requirements of the Hi3518 when ripples and noises occur. For details about the requirements on the power supply of each module, see the "Electrical Specifications" section in the *Hi3518 720p Processor Data Sheet*.



## 1.3 Design Recommendations for Peripheral Interfaces

### 1.3.1 USB Ports

#### USB Power Supply

The analog power AVDD33\_USB must be isolated from the digital power. You are advised to use planes to reduce the parasitic effect, decoupling noise, and power supply impedance. In addition, filtering capacitors are placed close to pins.

The digital power DVDD12\_USB and digital GND DVSS\_USB must not be interfered. Ensure that short and wide traces are used.

#### USB Protective Circuit

A protective circuit must be designed on the USB circuit to ensure electrostatic discharge (ESD) protection. To prevent the USB routing signals from being affected by protective components, design the PCB according to the following guidelines:

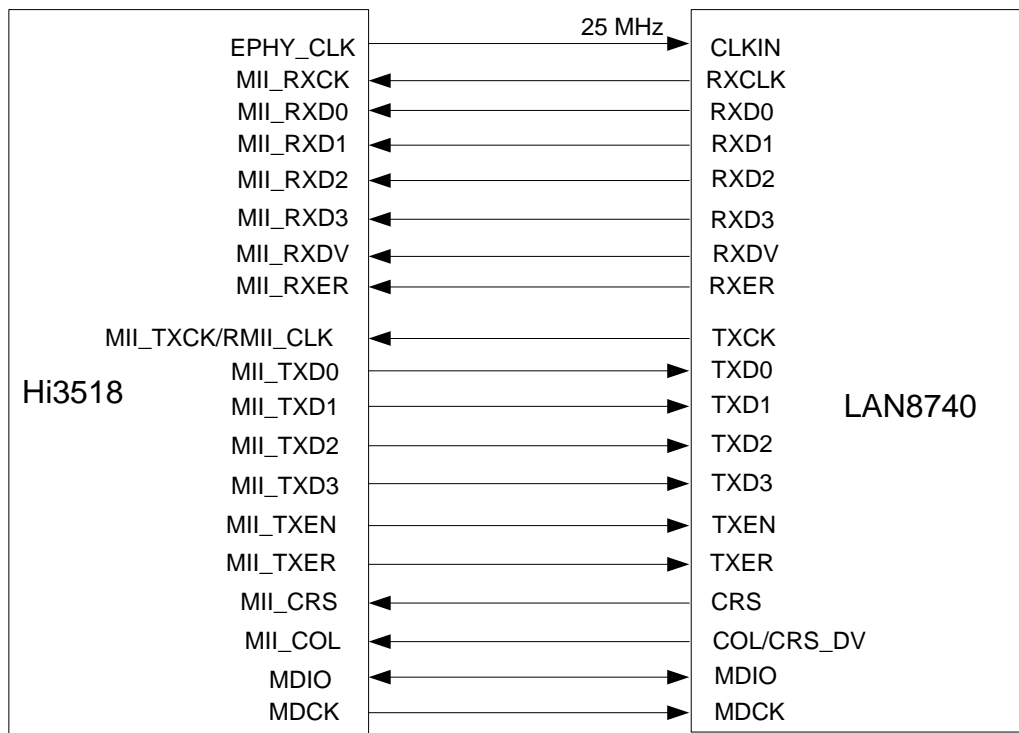
- Place the protective components close to the USB connector port.
- Use the TVS pipes with low parasitic capacitors as protective components, and ensure that the breakdown voltage is 8 kV, and the breakdown time is less than 1 ns.
- Ensure that the parasitic capacitors of the protective components connected to the high-speed USB 2.0 port are less than 1 pF.

### 1.3.2 MAC Interfaces

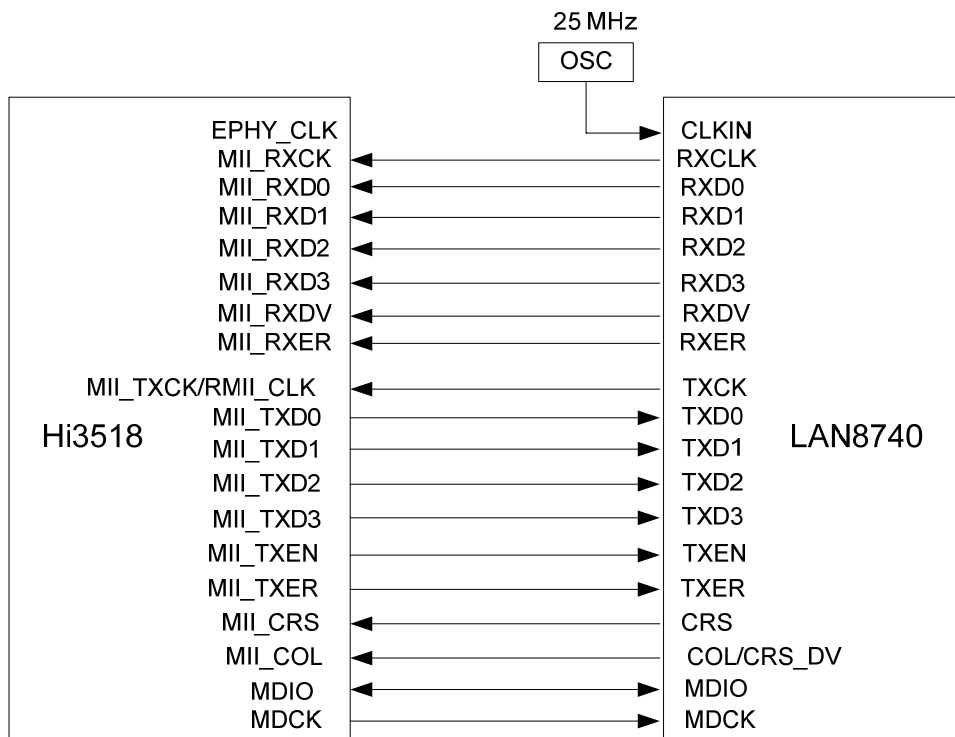
The MAC interfaces support the reduced media-independent interface (RMII) and media independent interface (MII) modes. For details about the signal connections in the two modes, see [Figure 1-12](#) to [Figure 1-15](#). The following uses LAN8740 (Ethernet PHY) as an example.



**Figure 1-12** Signal connections in MII mode (clocks provided by the Hi3518)

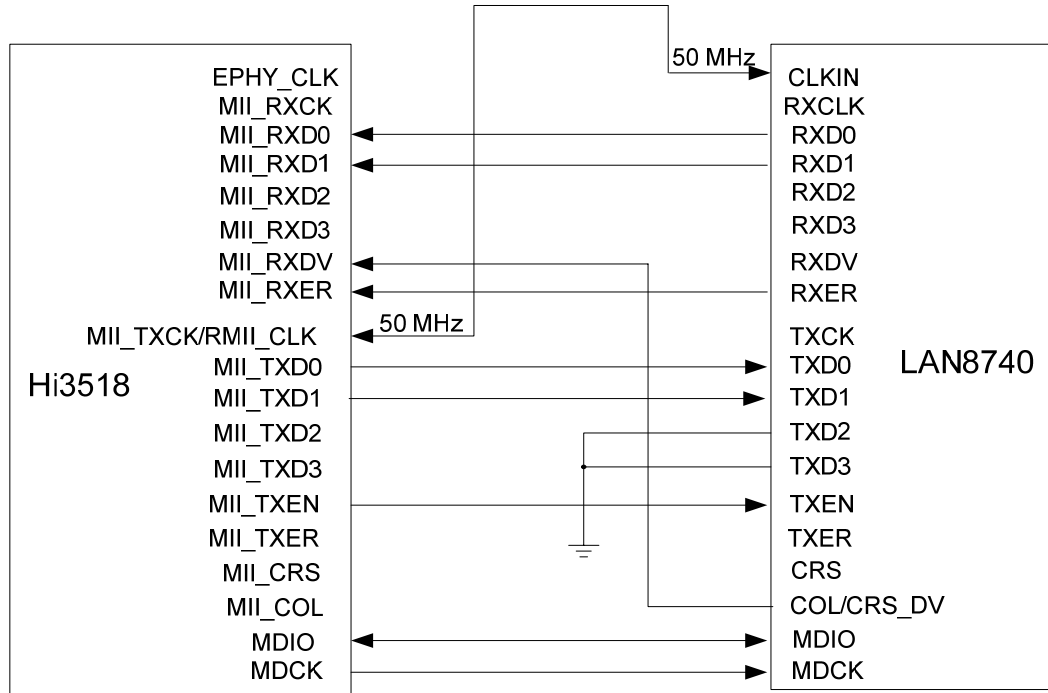


**Figure 1-13** Signal connections in MII mode (clocks provided by external components)





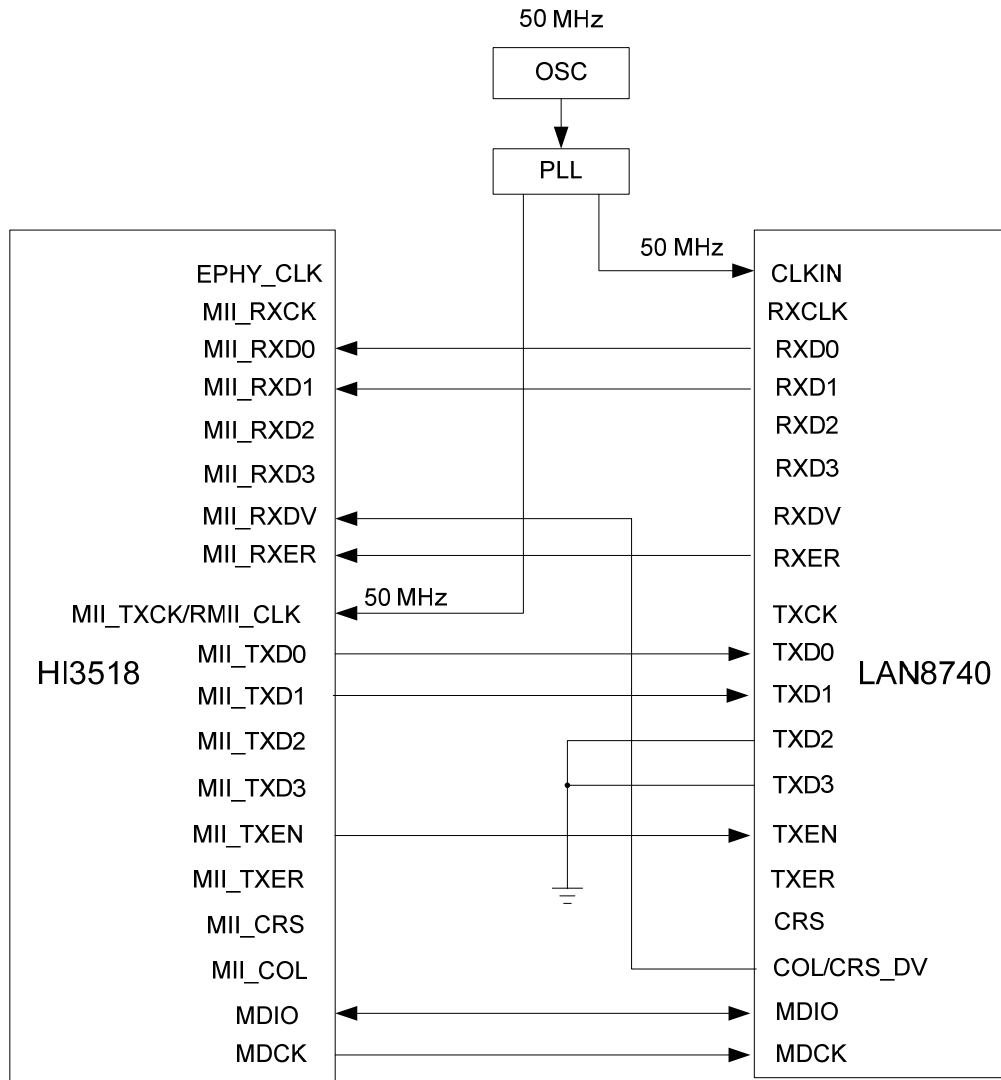
**Figure 1-14** Signal connections in RMI mode (clocks provided by the Hi3518)







**Figure 1-15** Signal connections in RMI mode (clocks provided by external components)



All MAC signals are connected in point-to-point topology. It is recommended that each PCB trace is less than or equal to 6 inches. The following are recommendations for designing matched resistors:

- Connect a 4.7 k $\Omega$  pull-up resistor to the MDIO signal.
- Connect a 22  $\Omega$  resistor to the MDCK signal in series at the source end to ensure signal quality.
- Connect the signals TXD0–TXD3 directly to the Hi3518.
- Connect a 33  $\Omega$  resistor in series to the TXCLK signal close to the PHY chip. In RMI mode, the TXCK signal is multiplexed as the RMII\_CLK signal, and a 33  $\Omega$  resistor is connected in series at the Hi3518 end.
- Connect the signals RXD0–RXD3 directly to the Hi3518.
- Connect a 33  $\Omega$  resistor in series to the RXCK signal close to the PHY chip.



## 1.3.3 Audio and Video Interfaces

### Analog DAC Interface

The Hi3518A provides a video digital-to-analog converter (VDAC), whereas the Hi3518C does not provide a VDAC.

- The VDAC\_IOUT pin is a composite video broadcast signal (CVBS) output pin. It must connect to an external 75  $\Omega$  resistor and then to GND. For details, see the schematic diagram of the Hi3518A demo board.
- The 3.3 V power supply for the VDAC must be isolated from the 3.3 V system power supply. Ensure that sufficient filtering capacitors connect to 3.3 V power pins. For details, see the schematic diagram of the Hi3518A demo board.
- The Hi3518 supports automatic CVBS detection. When no CVBS monitor is connected, the Hi3518 automatically disables the VDAC to reduce power consumption. If automatic CVBS detection is required, do not design the video buffer. If isolation and ESD are essential, the video buffer is recommended.

### Analog Audio Interface

The Hi3518A provides stereo-channel input and output interfaces: AC\_LINEL, AC\_LINER, AC\_OUTL, and AC\_OUTR. The Hi3518C provides mono-channel input and output interfaces: AC\_OUTL and AC\_LINEL.

- The 3.3 V power for the audio DAC must be isolated from the 3.3 V system power. The filtering capacitor connected to the AC\_VREF pin must be greater than or equal to 10  $\mu\text{F}$ . For details, see the schematic diagram of the Hi3518A demo board.
- AC\_LINEL and AC\_LINER can act as line or MIC input channels. If the input device is a passive MIC, connect the input signal to a resistor and then to the MIC\_BIAS pin. If the input device is an active line device such as the PC, the input signal does not need to connect to the MIC\_BIAS pin.
- Three series of sampling rates are supported: 32 kHz, 44.1 kHz, and 48 kHz.
  - The 32 kHz series sampling rates include 8 kHz, 16 kHz, 32 kHz, 64 kHz, and 128 kHz.
  - The 44.1 kHz series sampling rates include 11.025 kHz, 22.05 kHz, 44.1 kHz, 88.2 kHz, and 176.4 kHz.
  - The 48 kHz series sampling rates include 12 kHz, 24 kHz, 48 kHz, 96 kHz, and 192 kHz.
- The amplitude of the audio DAC is 0.875 V<sub>rms</sub> for full outputs.
- You are advised to connect an audio frequency amplifier and filtering circuit to the audio output pins AC\_OUTL and AC\_OUTR, which ensures excellent audio quality. For details, see the latest schematic diagram of the Hi3518PERB board.

### VI Interface

The Hi3518 has only one physical VI interface. This interface is called VI0.

- VI0 supports the raw data input, a maximum of 12-bit width, and a maximum of 74.25 MHz interface frequency.
- For a 12-bit sensor, the raw data pins connect to D0–D11 of VI0 in sequence.



- For a 10-bit sensor, the raw data pins connect to any 10-bit pins of VI0 in sequence. The pins D2 to D11 are recommended. Connect any unused pin to a 4.7 k $\Omega$  resistor and then to GND.

## VO Interface

The Hi3518 has only one physical video output (VO) interface. This interface is called VOU interface.

- If the VOU interface acts as the BT.1120 output, the bit width is 16 bits, and the maximum interface frequency is 74.25 MHz.
- The VOU interface does not provide external row sync and field sync signals and supports only internal synchronization.
- The VOU interface signal is multiplexed with the Ethernet (ETH) interface signal. If the ETH interface is used, the VOU interface is unavailable.
- For a BT.1120 signal, the upper eight bits are the Y (luminance) signal, and the lower eight bits are the C (chrominance) signal.

### 1.3.4 RTC Interface

The timing accuracy of the embedded RTC is affected by the manufacturing accuracy and temperature offset of the external crystal oscillator. For details, see the *RTC Correction Scheme Application Notes*.

If high timing accuracy is required, you are advised to select the external RTC with an embedded crystal oscillator or a crystal oscillator with the temperature compensation function.

## 1.4 Unused Pins

Do as follows if some pins are not used:

- If the JTAG function is not used, connect the JTAG\_EN pin to a 4.7 k $\Omega$  pull-down resistor, set the functional mode of the JTAG\_TCK, JTAG\_TMS, JTAG\_TRSTN, JTAG\_TDO, and JTAG\_TDI pins to GPIO, and set pin directions to output. No external processing is required.
- If the master chip is reset by the internal POR module, connect the POR\_SEL pin to a pull-down resistor. The RSTN pin can be floated.
- If the NAND flash is not used, set the functional modes of corresponding pins to GPIO and set pin directions to output. No external processing is required.
- If SPI0 and SPI1 are not used, set the functional modes of related pins to GPIO and set pin directions to output. No external processing is required.
- If the SAR\_ADC, VDAC, audio CODEC, and USB module are not used, retain their power supplies.
- If over-current protection is not required for the USB module, set the functional mode of the USB\_OVRCUR pin to GPIO and set the pin direction to output. No external processing is required.



## 1.5 Sensor Board

The following are design recommendations for the sensor board:

- To ensure picture quality, you are advised to use a low dropout (LDO) regulator to supply power to the sensor. Pay attention to the analog power and PLL power of the sensor. Typically, the core units of the sensor require a large current. The efficiency and heat dissipation of the LDO regulator need to be considered.
- The analog part and digital part of the sensor are separated. To be specific, the analog GND and digital GND are separated and connected by using  $0\ \Omega$  resistors in single-point mode. The analog power branch has an independent LDO regulator and does not share power branch with the digital power. During design of PCB layout and stacked architecture, ensure that the digital part and analog part do not intersect, avoiding interference and coupling.
- If the sensor board connects to the main board by using a connector, ensure that current return paths are provided for the data signals from the connector when specifying signals for connector pins. This avoids signal crosstalk. That is, each data signal trace or clock signal trace connects to an independent GND trace to form a current return path. Note that a clock signal trace is designed between two GND traces, and two data signal traces are designed between two GND traces. For details, see schematic diagrams.



# 2 PCB Design Recommendations

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## 2.1 PCB Design Recommendations for the Small System

### 2.1.1 Power Supplies of the Small System

#### Core Power Supply

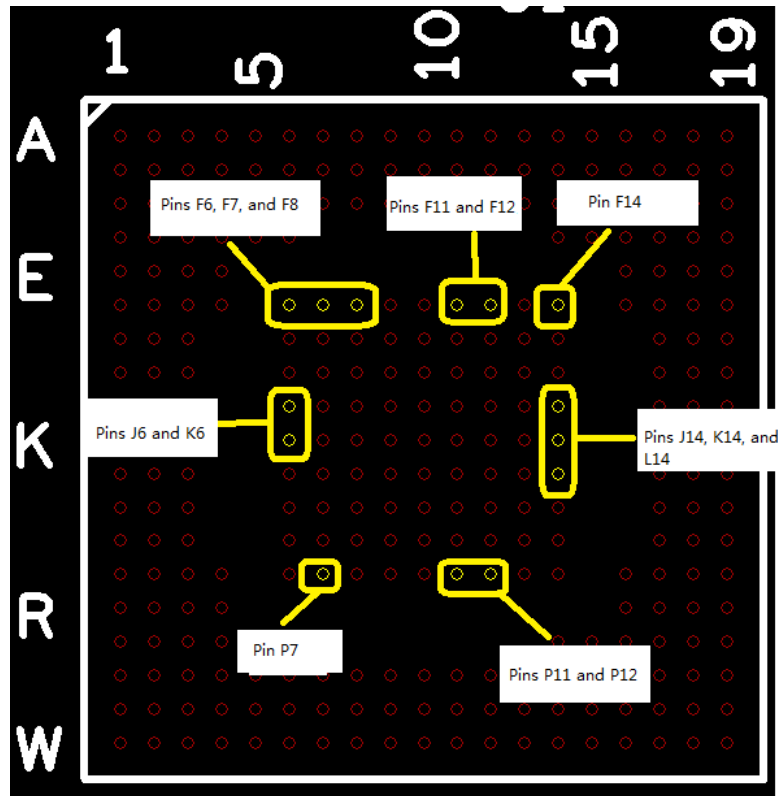
 **NOTE**

The current of the core power plane must be at least 1 A.

If the plane is used to supply power, 100 nF filtering capacitors must evenly connect to core power pins to reduce parasitic inductance. [Figure 2-1](#) shows the core power pins on the Hi3518A. Connect decoupling capacitors to these pins as follows:

- Ensure that capacitors are placed close to pins.
- Connect at least one 2.2  $\mu\text{F}$  capacitor to each of the pins F6, F7, and F8.
- Connect at least one 1  $\mu\text{F}$  capacitor to each of the pins J6 and K6.
- Connect at least one 10  $\mu\text{F}$  capacitor to each of the pins P11 and P12.
- Connect at least one 10  $\mu\text{F}$  capacitor to each of the pins J14, K14, and L14.
- Connect at least one 100  $\mu\text{F}$  capacitor to other pins each.

**Figure 2-1** Core power pins on the Hi3518A

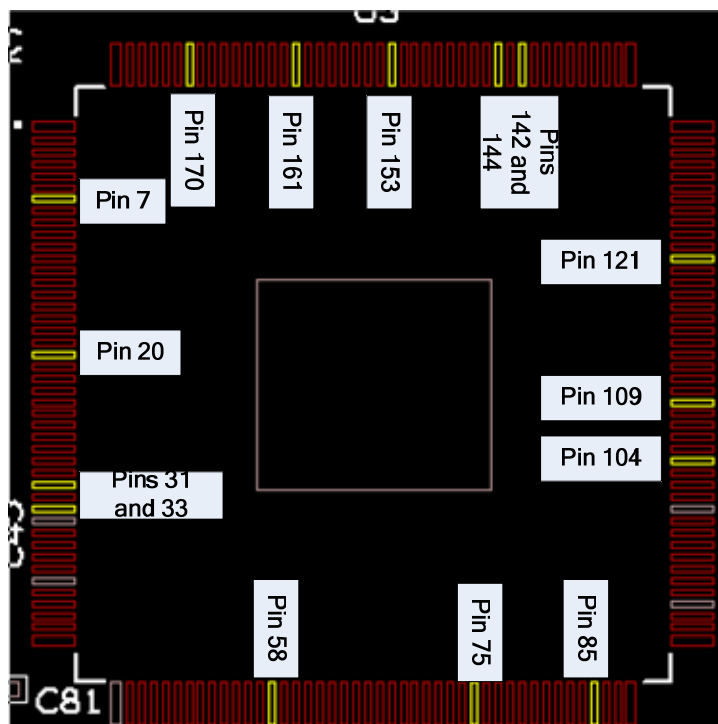


[Figure 2-2](#) shows the core power pins on the Hi3518C. Connect decoupling capacitors to these pins as follows:

- Ensure that capacitors are placed close to pins.
- Connect at least one 10  $\mu\text{F}$  capacitor to each of the pins P58 and P85.
- Connect at least one 10  $\mu\text{F}$  capacitor to each of the pins P161, P7, P153, and P104.
- Connect at least one 100 nF capacitor to other pins each.



**Figure 2-2** Core power pins on the Hi3518C



## DDR Power Supply

The 1.5 V/1.8 V power pins of the DDR2/DDR3 and the Hi3518 DDR need to be connected on the same power network. Decoupling capacitors need to be placed close to the power pins. At least one 10  $\mu$ F filtering capacitor connects to the Hi3518 DDR power and then to GND.

The VREF power supplies of the Hi3518 and DDR2/3 SDRAM (0.9 V for the DDR2 SDRAM and 0.75 V for the DDR3 SDRAM) must be isolated from other power supplies. You can use 20 mils or wider traces. Ensure that decoupling capacitors are placed close to each power pin and the VREF is shielded by surrounding ground traces.

The VREF design recommendations are as follows:

- Pay attention to the VREF routing when designing the VREF. According to the SSTL-15 standard, the noise of the VREF cannot be greater than  $\pm 1\%$  of the VREF level. To reduce the noise, ensure that the VREF traces are as wide as possible. In addition, you are advised to route the VREF traces at the power layer over a copper plane. This copper plane cannot be used as the reference plane for routing signal traces.
- A decoupling capacitor needs to connect to each VREF pin. The trace of each VREF pin must be as wide as possible, and the spacing between the trace and other signal traces must be 20–25 mils.

## I/O Power

Connect at least one 10  $\mu$ F filtering bypass capacitor to each DVDD33 pin and then to GND. Connect one 100 nF decoupling capacitor to each DVDD33 pin close to the power pin.



## 2.1.2 Clock and Reset Circuits

### Clock

The power and GND pins of the Hi3518 PLL unit include AVDD12\_PLL, AVDD33\_PLL, and AVSS\_PLL. Design the PCB according to the following guidelines:

- AVDD12\_PLL is the 1.2 V PLL power. You are advised to isolate it from the 1.2 V board digital power by using an EMI bead. The level deviation of the 1.2 V power must be within  $\pm 5\%$ .
- AVDD33\_PLL is the 3.3 V PLL power. You are advised to isolate it from the 3.3 V digital power by using an EMI bead. The level deviation of the 3.3 V power must be within  $\pm 5\%$ .
- AVSS\_PLL is the reference GND of the PLL circuit. The decoupling capacitors for AVDD12\_PLL and AVDD33\_PLL must be connected by referring to AVSS\_PLL as the GND plane. AVSS\_PLL must connect to the digital board GND through a single point.
- The traces of system clocks and crystal oscillator circuit must be as short as possible. The traces must be surrounded by GND traces.

### Reset

The Hi3518 can be reset internally or externally by configuring the POR\_SEL pin. During power-on, if POR\_SEL is 0, internal reset is enabled; if POR\_SEL is 1, internal reset is disabled, and the chip is reset externally. For details about the circuit design, see the schematic diagram of the Hi3518A demo board.

## 2.1.3 DDR Signals



### NOTE

In this section, the DDR signal design is described by using the 16-bit DDR as an example.

### Clock Signal CLK

The following are requirements on the trace length of the clock signal CLK:

- The trace length of the CLK signal must be shorter than or equal to 4.5 inches.
- The traces of CLK differential signals must be routed in differential mode, and the deviation of each pair of differential traces is  $\pm 5$  mils.

### DQS Signals

For DQS signals, the trace requirements are as follows:

- Each pair of DQS differential traces must have the same length, and the deviation of each pair of differential traces is less than 5 mils.
- DQS traces are routed based on the CLK trace length. The deviation of the DQS trace is  $\pm 250$  mils relative to the CLK trace length.

### Data Signals DQ[0:15]

The trace length of data signals DQ[15:0] is based on the DQS trace length.





- DQ[7:0] are routed based on the DQS 0 trace length, and the length deviation is  $\pm 200$  mils.
- DQ[15:8] are routed based on the DQS 1 trace length, and the length deviation is  $\pm 200$  mils.

## Data Mask Signals

The data mask (DM) signal traces are routed based on the DQS trace length. The requirements are as follows:

- The DM 0 trace is routed based on the DQS 0 trace length, and the length deviation is  $\pm 200$  mils.
- The DM 1 trace is routed based on the DQS 1 trace length, and the length deviation is  $\pm 200$  mils.

## Address Signals ADDR[0:14]

ADDR[0:14] traces are routed based on the CLK trace length, and the length deviation is  $\pm 700$  mils.

## Control Signals

The traces of the control signals including BA[0:2], DM, CKE, CSN, WEN, CASN, RASN, and ODT are routed based on the CLK trace length, and the length deviation is  $\pm 700$  mils.

## PCB Routing Recommendations

Route traces on a PCB according to the following guidelines:

- Route DDR3 SDRAM data traces by referring to the GND plane and avoid crossing the power and GND plane splits. Ensure that all DDR signal traces are routed by referring to a complete plane.
- Ensure that signal traces are as short as possible, and minimize the use of vias to ensure the impedance continuity of traces. The impedance of the single-end signal PCB trace is  $50 \Omega \pm 10\%$ , and the impedance of the differential PCB trace is  $100 \Omega \pm 10\%$ .
- Ensure that the spacing between adjacent traces meets the 3W rule. The 3W rule indicates that the trace spacing is three times of the trace width.
- The spacing between a DDR3 SDRAM signal trace and a non-DDR3 SDRAM signal trace must be at least 20 mils.
- Connect a decoupling capacitor to each VREF pin. The trace of each VREF pin must be as wide as possible, and the spacing between the VREF pin trace and other signal traces is 20–25 mils.
- The Hi3518 does not support DDR training.

### 2.1.4 SPI Flash

Route traces on a PCB according to the following guidelines:

- Ensure that the spacing between adjacent traces meets the 3W rule.
- Connect a  $33 \Omega$  resistor to the clock signal in series to ensure signal quality.



## 2.1.5 Ethernet Port Signal

- Ensure that the spacing between adjacent traces meets the 3W rule.
- Connect a 33  $\Omega$  resistor to the clock signal in series to ensure signal quality.

## 2.2 PCB Design Recommendations for Peripheral Interfaces

### 2.2.1 USB Port Design

#### USB PCB Design

Route the data signal traces of the USB 2.0 host port in differential mode to ensure signal quality. You are advised to design the PCB routing according to the following guidelines to meet the 480 MHz requirement for the USB 2.0 port:

- The differential data traces must be short and straight and the internal differential traces must have the same length. It is recommended that the trace deviation be within 5 mils.
- The impedance deviation of each USB differential trace is 90  $\Omega \pm 10\%$ .
- Route traces on the routing planes close to GND, and do not change the routing plane.
- Route differential data traces by referring to complete GND planes, and do not cross the plane splits.
- Minimize the use of vias and corners when routing differential data traces. When corners are required, use arcs or 135° turns but not a 90° turn. This reduces signal reflection and impedance variance.
- Route the differential data traces away from other high-speed cyclic signals and strong current signals, and ensure that the spacing is greater than 50 mils, reducing crosstalk. In addition, route the differential data traces away from low-speed non-cyclic signals and ensure that the spacing is at least 20 mils.
- Place the REXT resistor close to the Hi3518.

#### Power Supply of USB Unit

The Hi3518 has one USB host port. The power pins and GND pins of the USB unit are AVDD33\_USB, AVSS\_USB, and DVDD12\_USB that correspond to the 3.3 V power, 3.3 V GND, and 3.3 V core power respectively.

AVDD33\_USB is the 3.3 V analog power, and AVSS\_USB is the 3.3 V analog GND. The level deviation of AVDD33\_USB must be within  $\pm 7\%$ . You are advised to isolate the 3.3 V analog USB power from the 3.3 V digital board power by using an EMI bead, and place filtering capacitors close to AVDD33\_USB and AVSS\_USB pins.

#### Routing of the USB Protective Circuit

A USB protective circuit is required to implement electrostatic discharge (ESD) protection. To avoid the USB trace signals from being attenuated by the protective components, design the PCB according to the following guidelines:

- Place protective components close to the USB port.



- Use the TVS pipes with low parasitic capacitors as protective components.
- The parasitic capacitors connected to the protective components on the high-speed USB 2.0 port must be less than 2 pF.

## 2.2.2 Audio Circuit Design

### 2.2.2.1 DAC Interface

The Hi3518A provides a VDAC, whereas the Hi3518C does not provide a VDAC.

#### VDAC PCB Design

To ensure the output picture quality of the VDAC, design PCB routing according to the following guidelines:

- Connect the VDAC output pin to a resistor and then to GND. The resistor is placed close to the Hi3518A.
- Place the analog video filtering circuit close to the Hi3518A.
- Use a VDAC to drive only a single load. If you need to drive multiple loads to output multi-channel video signals, drives are recommended.
- The VDAC unit has a separate analog video GND plane. All analog signal traces are routed by referring to this plane.
- Route all analog signal traces on the planes close to the analog video GND plane, and do not change the routing plane.
- Do not route other signal traces by referring to the analog video GND plane.
- When corners are required, use arcs or 45° turns but not a 90° turn to reduce signal reflection.

#### Power Supply of the VDAC Unit

AVDD3V3\_VDAC of the VDAC unit is an analog video power pin. Design the PCB according to the following guidelines:

- Use a GND plane as the analog video GND and digital GND.
- The deviation of the 3.3 V analog video power must be within  $\pm 5\%$ .
- Isolate the 3.3 V analog VDAC power from the 3.3 V digital board power by using an EMI bead, and place filtering capacitors close to the AVDD33\_VDAC pin.
- Connect the VDAC\_COMP to a 10  $\mu$  capacitor and a 10 nF capacitor in parallel, and then to the AVDD33\_VADC pin.

#### Protective Circuit of the VDAC Port

A protective circuit for the VDAC output port is required to implement ESD protection. You are advised to design the PCB according to the following guidelines:

- Use the video buffer circuit for video output to separate the video port of the Hi3518 from the add-on port.
- Place the protective components close to the connector of the VDAC output port.
- Use TVS pipes or switch diodes as protective components.



## 2.2.2.2 VI and VO Interfaces

The Hi3518 has a VO interface and a VI interface. The following are routing requirements:

- The VO data signal trace is routed based on the CLK signal trace. The deviation length is less than or equal to 1000 mils.
- Route traces of the VI data signal, row sync signal, and field sync signal based on the CLK signal trace. The deviation length is less than or equal to 1000 mils.
- The VI interface has 1.8 V and 3.3 V power domains. Route the traces of the VI data signal, row sync signal, field sync signal, and SPI 0 at the top layer by referring to the GND plane. This avoids sharp changes of characteristic impedance when signal traces are routed by referring to the 1.8 V or 3.3 V power plane.
- If the 1.8 V power domain is selected for the VI interface, the signal trace must be less than or equal to 4 inches.



# 3 Heat Design Recommendations

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## 3.1 Working Condition

For details about power supply parameters, temperature parameters, and thermal resistance parameters for the Hi3518, see section 2.6 "Electrical Specifications" in the *Hi3518 HD IP Camera SoC Data Sheet*.

## 3.2 Reference Design for Circuit Heat Dissipation

### 3.2.1 Schematic Diagram

#### Power Supply

Ensure the efficiency of the board power tree as long as the power supply is stable. That is, you design the board power supply optimally and use few large-voltage-difference LDO components. This reduces the heat produced during power supply conversion.

The board power supplies for peripherals such as the secure digital (SD) card and USB device can be reserved during board design. These power supplies can be shut down when they are not used. The main integrated circuits (ICs) of the board must support the power-down mode.

#### Low-Power Configurations for Idle Modules

Modules such as the VDAC, USB module and NANDC may not be used in actual scenarios. You can set the modes of these modules to power-down mode or default mode.

### 3.2.2 PCB

#### Component Layout

Lay out components based on the product architecture and heat dissipation design:

- Place the components that consume much power and generate much heat in a distributed manner to avoid overheating of some parts and ensure the reliability and efficiency of components. You are advised to place the Hi3518 away from power supplies.



- Design the product architecture optimally to ensure that the heat produced internally can be dissipated.

## Routing

The routing recommendations are as follows:

- For the connection style of the vias under the Hi3518, select the full connection style but not the thermal connection style to improve the board dissipation efficiency.
- Connect the GND signals and 1.2 V, 1.5 V (or 1.8 V), and 3.3 V power signals over copper planes. When the signal over-current performance is ensured, you are advised to punch more vias on copper planes.
- Increase the size of copper planes under and around the components that produce much heat to ensure that the heat of the PCB can be dissipated effectively. For the inductors and power chips, place them in a distributed manner and increase the size of copper planes around them.