

NOT RECOMMENDED FOR NEW DESIGN CONTACT US



AP1539

18V 4A 300KHz BUCK CONVERTER

Description

AP1539 consists of step-down switching regulator with PWM control. These devices include a reference voltage source, oscillation circuit, error amplifier, internal PMOS.

AP1539 provides low-ripple power, high efficiency, and excellent transient characteristics. The PWM control circuit is able to vary the duty ratio linearly from 0 up to 99%. This converter also contains an error amplifier circuit as well as a soft-start circuit that prevents overshoot at startup. An enable function, an over current protect function and a short circuit protect function are built inside, and when OCP or SCP happens, the operation frequency will be reduced from 300KHz to 50KHz. Also, an internal compensation block is built in to minimum external component count. With the addition of an internal Pchannel Power MOS, a coil, capacitors, and a diode connected externally, these ICs can function as step-down switching regulators. They serve as ideal power supply units for portable devices when coupled with the SOP-8L-DEP mini-package, providing such outstanding features as low current consumption. Since this converter can accommodate an input voltage up to 18V, it is also suitable for the operation via an AC adapter.

Features

Input voltage: 3.6V to 18VOutput voltage: 0.8V to V_{CC}

Output current: up to 4A

Duty ratio: 0% to 99% PWM control

Oscillation frequency: 300KHz typ.

· Soft-start like, Current limit and Enable function

• Thermal Shutdown function

Built-in internal SW P-channel MOS

• SOP-8L-DEP: Available in "Green" Molding Compound (No Br, Sb)

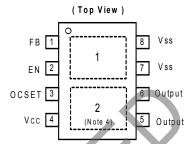
Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)

• Halogen and Antimony Free. "Green" Device (Note 3)

 For automotive applications requiring specific change control (i.e. parts qualified to AEC-Q100/101/200, PPAP capable, and manufactured in IATF 16949 certified facilities), please contact us or your local Diodes representative.

https://www.diodes.com/quality/product-definitions/

Pin Assignments



SOP-8L-DEP (Dual Exposed Pads)

Note: 4. Exposed pad 1 is connected to V_{SS} and exposed pad 2 is connected to Output. The board layout for exposed pads needs to be considered to avoid short circuit.

Applications

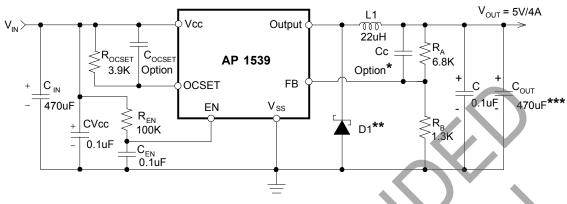
- PC Motherboard
- LCD Monitor
- Graphic Card
- DVD-Video Player
- Telecom Equipment
 - ADSL Modem
- Printer and other Peripheral Equipment
- Microprocessor core supply

Notes:

- 1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.
- 2. See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
- 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.



Typical Application Circuit



 $V_{OUT} = V_{FB} \times (1 + R_A/R_B)$ $R_B = 0.7K\sim5K$ ohm

Notes:

- * Typical feedback compensation (Cc): 5600pF.

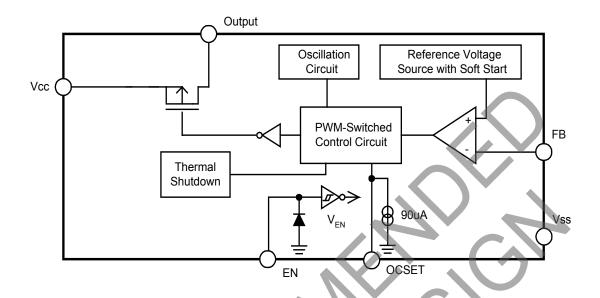
 ** Suggested Diodes Incorporated Power Schottky P/N: B540 series or PDS540.
- *** Suggested C_{OUT} for V_{OUT} < 1V; 680uF.

Pin Descriptions

Pin Name	Pin No.	Description
FB	1	Feedback pin
EN	2	Power-off pin H. Normal operation (Step-down operation) L: Step-down operation stopped (All circuits deactivated)
OCSET	3	Add an external resistor to set max output current
V _{CC}	4	IC power supply pin
Output	5, 6	Switch Pin. Connect external inductor/diode here. Minimize trace area at this pin to reduce EMI
Vss	7, 8	GND Pin



Block Diagram



Absolute Maximum Ratings

Symbol	Parameter	Rating	Unit
ESD HBM	Human Body Model ESD Protection	5.5	KV
ESD MM	Machine Model ESD Protection	200	V
V _{CC}	V _{CC} Pin Voltage	V _{SS} - 0.3 to V _{SS} + 20	V
V _{FB}	Feedback Pin Voltage	V _{SS} - 0.3 to V _{CC}	V
V _{EN}	EN Pin Voltage	V _{SS} - 0.3 to V _{CC}	V
V _{OUT}	Switch Pin Voltage	V_{SS} - 0.3 to V_{CC}	V
P _D	Power Dissipation	Internally limited	mW
T _J	Operating Junction Temperature Range	-40 to +125	°C
T _{ST}	Storage Temperature Range	-65 to +150	°C

Caution: The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

Recommended Operating Conditions

Symbol	Parameter	Min	Max	Unit
V_{IN}	Input Voltage (Note 5)	3.6	18	V
lout	Output Current	0	4	Α
T _A	Operating Ambient Temperature	-20	+85	°C

Note: 5. For the operations in low input voltage, AP1539 can tolerate down to 3.6V but max output current loading will be less than 4A. For nominal applications in such low input voltage range, especially lower than 4V, a higher ROCSET with larger heat sink is recommended.

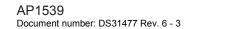


Electrical Characteristics

(V_{IN} = 12V, T_A = +25°C, unless otherwise specified)

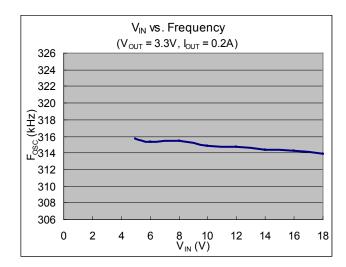
Symbol	Parameter	Conditions	Min	Тур.	Max	Unit	
V _{FB}	Feedback Voltage	I _{OUT} = 0.1A	0.784	0.8	0.816	V	
I _{FB}	Feedback Bias Current	I _{OUT} = 0.1A	_	0.1	0.5	μΑ	
I _{SHDN}	Current Consumption During Power Off	V _{EN} = 0V	_	10	1	μΑ	
ΔV_{OUT} / V_{IN}	Line Regulation	V _{IN} = 5V~18V	_	1	2	%	
ΔV _{OUT} / V _{OUT}	Load Regulation	I _{OUT} = 0.1 to 4A	_	0.2	0.5	%	
fosc	Oscillation Frequency	Measure waveform at SW pin	240	300	400	KHz	
f _{OSC1}	Frequency of Current Limit or Short Circuit Protection	Measure waveform at SW pin		50		KHz	
V_{IH}	EN Pin Input Voltage	Evaluate oscillation at SW pin	2.0			>	
V_{IL}	LIVI III IIIput Voltage	Evaluate oscillation stop at SW pin		70	0.8	· ·	
I_{SH}	EN Pin Input Leakage Current	EN Pin High	_	20	F	μΑ	
I_{SL}		EN Pin Low	<u> </u>	-10	<u> </u>	μΑ	
I OCSET	OCSET Pin Bias Current	_	75	90	105	μΑ	
Dagger	Internal MOSFET R _{DS(ON)}	$V_{IN} = 5V$, $V_{FB} = 0V$		90	-	mΩ	
R _{DS(ON)}		$V_{IN} = 12V, V_{FB} = 0V$		50			
EFFI	Efficiency	V _{IN} = 12V, V _{OUT} = 5V I _{OUT} = 4A		92	-	%	
T _{SHDN}	Thermal shutdown threshold	- () ()		+150		°C	
T _{HYS}	Thermal shutdown hysteresis	- 617		+55	_	°C	
θјс	Thermal Resistance Junction-to-Case	SOP-8L-DEP (Note 6)	_	26	_	°C/W	

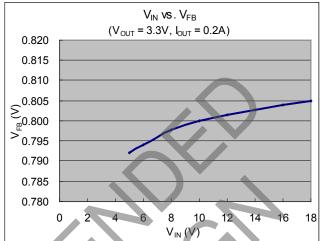
Note: 6. Test condition for SOP-8L-DEP: Devices mounted on 2oz copper, minimum recommended pad layout on top & bottom layer with thermal vias, double sided FR-4 PCB.

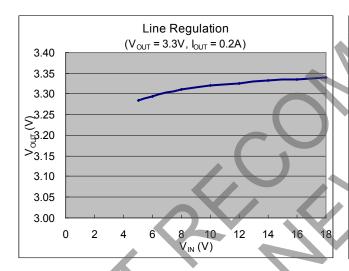


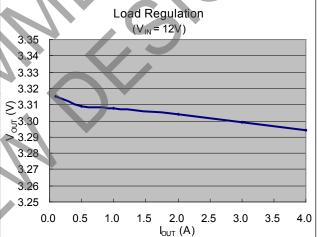


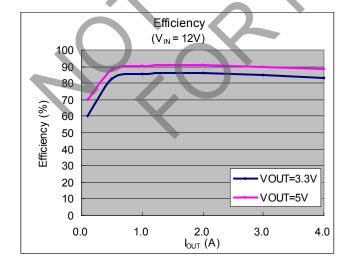
Typical Performance Characteristics

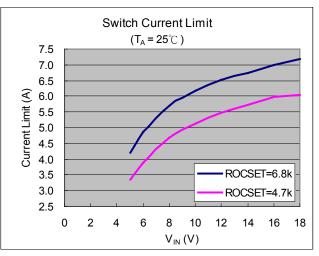






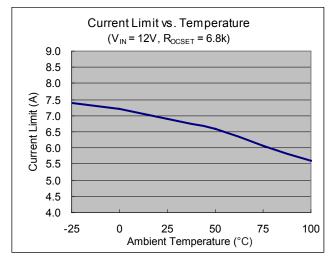


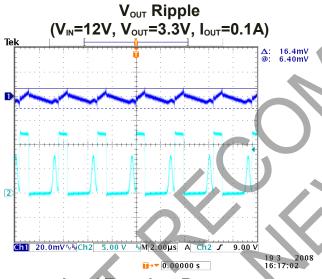


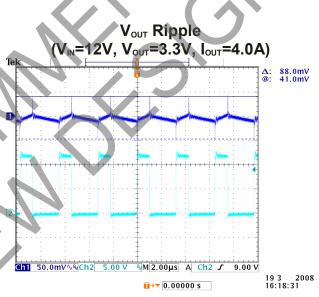


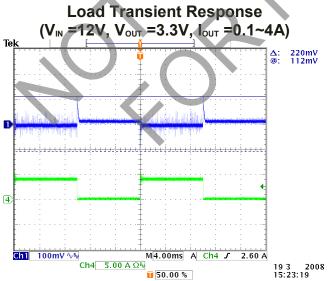


Typical Performance Characteristics (continued)



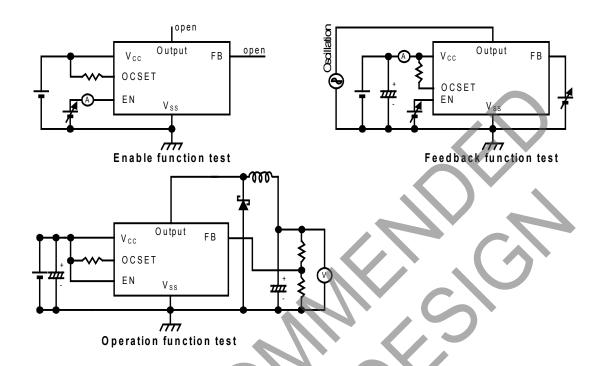








Test Circuit



Functional Description

PWM Control

The AP1539 is a DC/DC converter that employs pulse width modulation (PWM) scheme. Its pulse width varies in the range of 0% to 99%, based on the output current loading. The output ripple voltage caused by the PWM high frequency switching can easily be reduced through an output filter. Therefore, this converter provides a low ripple output supply over a broad range of input voltage & output current loading.

Under Voltage Lockout

The under voltage lockout circuit of the AP1539 assures that the high-side MOSFET driver remains in the off state whenever the supply voltage drops below 3.3V. Normal operation resumes once V_{CC} rises above 3.5V.

Current Limit Protection

The current limit threshold is set by external resistor R_{OCSET} connected from V_{CC} supply to OCSET pin. The internal sink current I_{OCSET} (90uA typical) across this resistor sets the voltage at OCSET pin. When the PWM voltage is less than the voltage at OCSET, an over-current condition is triggered.

The current limit threshold is given by the following equation:

$$I_{PEAK} \times R_{DS(ON)} = I_{OCSET} \times R_{OCSET}$$

$$I_{PEAK} > I_{OUT(MAX)} + \frac{(\Delta I)}{2}$$

where,

$$\Delta I = \frac{V_{IN} - V_{OUT}}{fs \times L} \times \frac{V_{OUT}}{V_{IN}}$$

 I_{PEAK} is the output peak current; $R_{DS\;(ON)}$ is the MOSFET ON resistance; F_S is the PWM frequency (300KHz typical). Also, the inductor value will affect the ripple current ΔI .



Functional Description (continued)

The above equation is recommended for input voltage range of 5V to 18V. For input voltage lower than 5V, higher than 18V or ambient temperature over +100°C, higher Rocset is recommended.

The recommended minimum Rocset value is summarized below:

Vout	V _{IN} (V)			
(V)	5V	12V	18V	
0.8	6.8K	3.9K	4.7K	
1.0	6.8K	3.9K	4.7K	
1.2	6.8K	3.9K	4.7K	
1.8	6.8K	3.9K	4.7K	
2.5	6.8K	3.9K	4.7K	
3.3	6.8K	3.9K	4.7K	
5.0	N/A	3.9K	5.6K	

Inductor Selection

For most designs, the operates with inductors of 22µH to 33µH. The inductor value can be derived from the following equation:

$$L = \frac{V_{IN} - V_{OUT}}{fs \times \Delta I} \times \frac{V_{OUT}}{V_{IN}}$$

Where ΔI_L is inductor Ripple Current. Large value inductors lower ripple current and small value inductors result in high ripple current. Choose inductor ripple current approximately 15% of the maximum load current 4A, ΔI_L =0.6A. The DC current rating of the inductor should be at least equal to the maximum load current plus half the ripple current to prevent core saturation (4A+0.3A).

Input Capacitor Selection

This capacitor should be located close to the IC using short leads and the voltage rating should be approximately 1.5 times the maximum input voltage. The RMS current rating requirement for the input capacitor of a buck regulator is approximately 1/2 the DC load current. A low ESR input capacitor sized for maximum RMS current must be used. A 470µF low ESR capacitor for most applications is sufficient.

Output Capacitor Selection

The output capacitor is required to filter the output voltage and provides regulator loop stability. The important capacitor parameters are the 100KHz Equivalent Series Resistance (ESR), the RMS ripples current rating, voltage rating and capacitance value. For the output capacitor, the ESR value is the most important parameter. The output ripple can be calculated from the following formula.

$$V_{RIPPLE} = \Delta I_{L} \times ESR$$

The bulk capacitor's ESR will determine the output ripple voltage and the initial voltage drop after a high slew-rate transient.

An aluminum electrolytic capacitor's ESR value is related to the capacitance and its voltage rating. In most case, higher voltage electrolytic capacitors have lower ESR values. Most of the time, capacitors with much higher voltage ratings may be needed to provide the low ESR values required for low output ripple voltage.

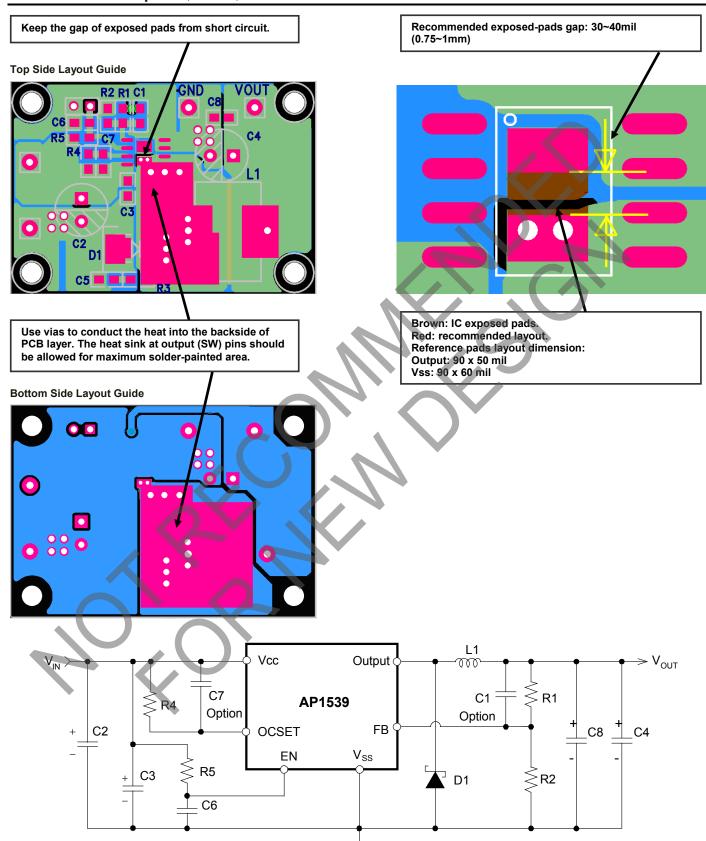
PCB Layout Guide

If you need low T_C & T_J or large PD (Power Dissipation), The dual SW pins(5& 6) and V_{SS} pins(7& 8)on the SOP-8L package are internally connected to die pad, The evaluation board should be allowed for maximum copper area at output (SW) pins.

- 1. Connect FB circuits (R₁, R₂, C₁) as closely as possible and keep away from inductor flux for pure V_{FB}.
- 2. Connect C3 to V_{CC} and V_{SS} pin as closely as possible to get good power filter effect.
- 3. Connect R4 to V_{CC} and OCSET pin as closely as possible.
- 4. Connect ground side of the C2 & D1 & C4 as closely as possible and use ground plane for best performance.

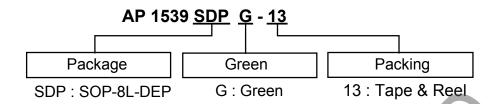


Functional Description (continued)





Ordering Information

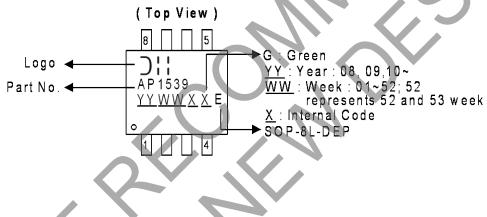


Device	Package Packaging		13" Tape and Reel		
Device	Code	(Note 7)	Quantity	Part Number Suffix	
AP1539SDPG-13	SDP	SOP-8L-DEP	2500/Tape & Reel	-13	

Note: 7. Pad layout as shown as shown in Diodes Incorporated's package outline PDFs, which can be found on our website at http://www.diodes.com/package-outlines.html.

Marking Information

(1) SOP-8L-DEP

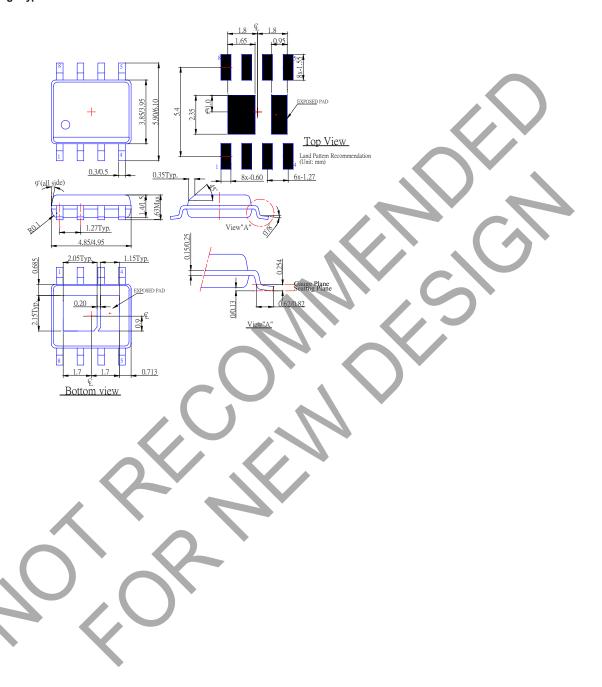




Package Outline Dimensions (All Dimensions in mm)

Please see http://www.diodes.com/package-outlines.html for the latest version.

(1) Package type: SOP-8L-DEP





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