

Introduction to VNA Basics

PRIMER



The Vector network analyzer or VNA is an important test instrument that has helped make countless modern wireless technologies possible. Today, VNAs are used in a wide range of RF and high frequency applications. In design applications, simulations are used to accelerate time-to-market by reducing physical prototype iterations. VNAs are used to validate these design simulations. In manufacturing applications, RF components or devices are assembled and tested based on a certain set of specifications. VNAs are used to quickly and accurately validate the performance of these RF components and devices.

This paper discusses why VNAs are used and how they are unique compared to other RF test equipment. We'll define S-Parameters, the fundamental VNA measurement, and how best to use them when evaluating your Device-Under-Test or DUT. We'll review various VNA calibration techniques and show how VNA user calibrations help achieve the best accuracy possible. Finally, we'll review typical VNA measurements such as swept frequency measurements, time domain measurements, and swept power measurements and how they're used and why they are important.

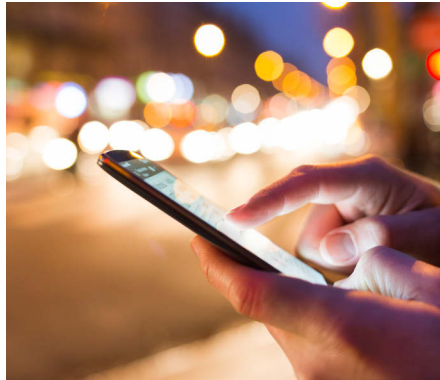
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Not for measuring
WiFi networks



Not for drive testing
mobile phone networks



Not for computer
networks or clouds



FIGURE 1. Today there are a wide variety of networks, each with its own network analyzer. The vector network analyzer, discussed in this document, is used for a different kind of network and was defined long before any of these networks existed.

Vector Network Analyzer Overview

Today, the term “network analyzer”, is used to describe tools for a variety of “networks” (Figure 1). For instance, most people today have a cellular or mobile phone that runs on a 3G or 4G “network”. In addition, most of our homes, offices and commercial venues all have Wi-Fi, or wireless LAN “networks”. Furthermore, many computers and servers are setup in “networks” that are all linked together to the cloud. For each of these “networks”, there exists a certain network analyzer tool used to verify performance, map coverage zones and identify problem areas.

However, the network analyzer of interest in this paper is used for a different kind of network and was defined long before any of these networks existed. The first VNA was invented around 1950 and was defined as an instrument that measures the network parameters of electrical networks (Figure 2). In fact, it can be said that the VNA has been used over the years to help make all the networks mentioned above possible. From mobile phone networks, to Wi-Fi networks, to computer networks and the to the cloud, all of the most common technological networks of today were made possible using the VNA that was first invented over 60 years ago.

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FIGURE 2. Vector Network Analyzers or VNAs were invented in the 1950s and are actively used around the world today.

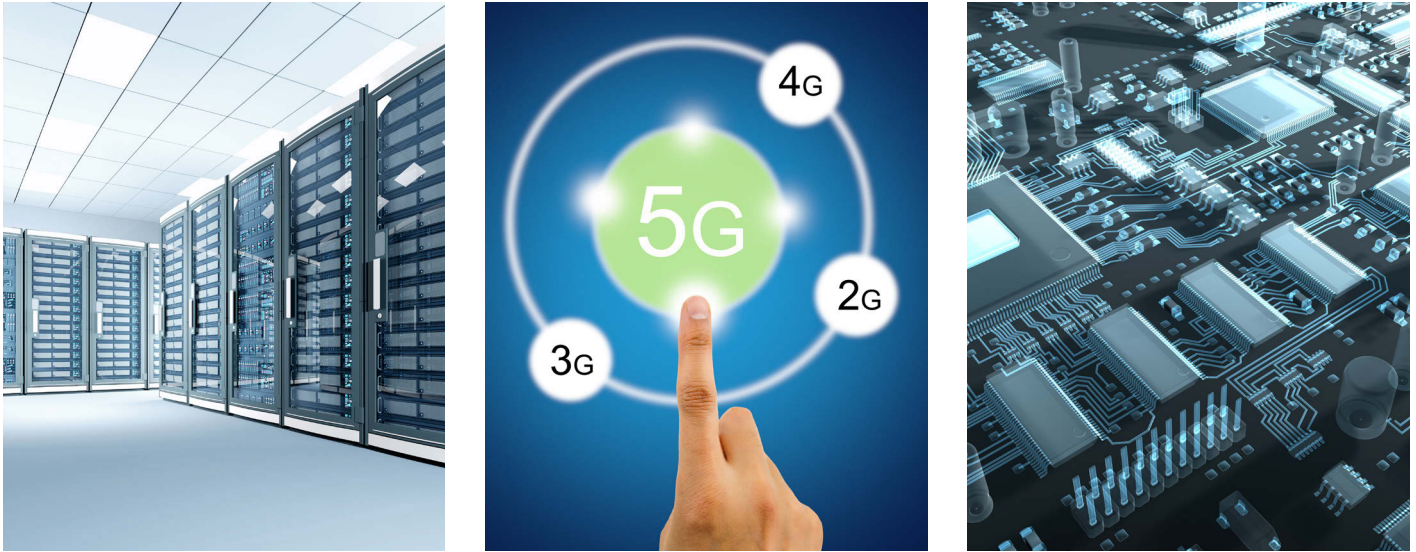


FIGURE 3. VNAs are used to make most modern technologies possible.

WHO NEEDS A VNA

All wireless solutions have transmitters and receivers, and each contains many RF and microwave components. This includes not only smartphones and WiFi networks, but also connected cars and IoT (Internet of Things) devices. Additionally, computer networks today operate at such high frequencies that they are passing signals at RF and microwave frequencies. Figure 3 shows a range of example applications that exist today with the help of VNAs.

VNAs are used to test component specifications and verify design simulations to make sure systems and their components work properly together. R&D engineers and manufacturing test engineers commonly use VNAs at various stages of product development. Component designers need to verify the performance of their components such as amplifiers, filters, antennas, cables, mixers, etc. The system designer needs to verify their component specs to ensure that the system performance they're counting on meets their subsystem and system specifications. Manufacturing lines use VNAs to make sure that all products meet specifications before they're shipped out for use by their customers. In some cases, VNAs are even used in field operations to verify and troubleshoot deployed RF and microwave systems.

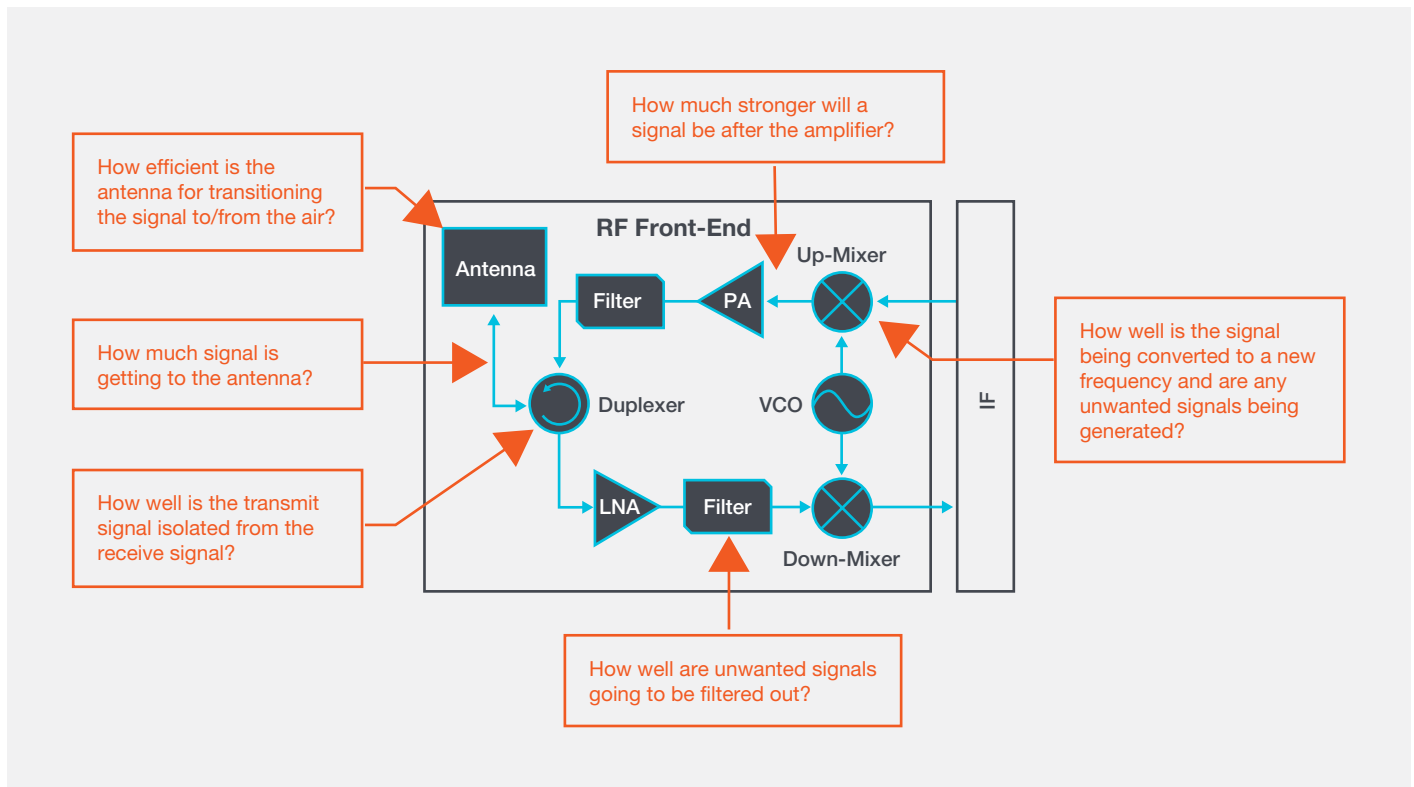


FIGURE 4. VNAs may be used to verify component, subsystem and system level performance.

As an example, Figure 4 shows an RF system front end and how different components and parts of the system are tested with a VNA. For the antenna, it is important to understand how efficient the antenna is at transitioning the signal to and from the air. As we'll explain later, this is determined by using a VNA to measure the return loss or VSWR of the antenna.

Looking at the right side of Figure 4, the up-mixer takes the IF signal and mixes it with an oscillator (VCO) to produce the RF signal. How well is the signal being converted to a new frequency? Are any unwanted signals being generated? What power levels are the most efficient at driving the mixer? VNAs are used to answer these questions.

From a system design point of view, how much signal goes through the RF board and out of the antenna? On the receive side, how effective is the duplexer in providing isolation between the transmit and the receive signal? All of these questions can be answered using a VNA.

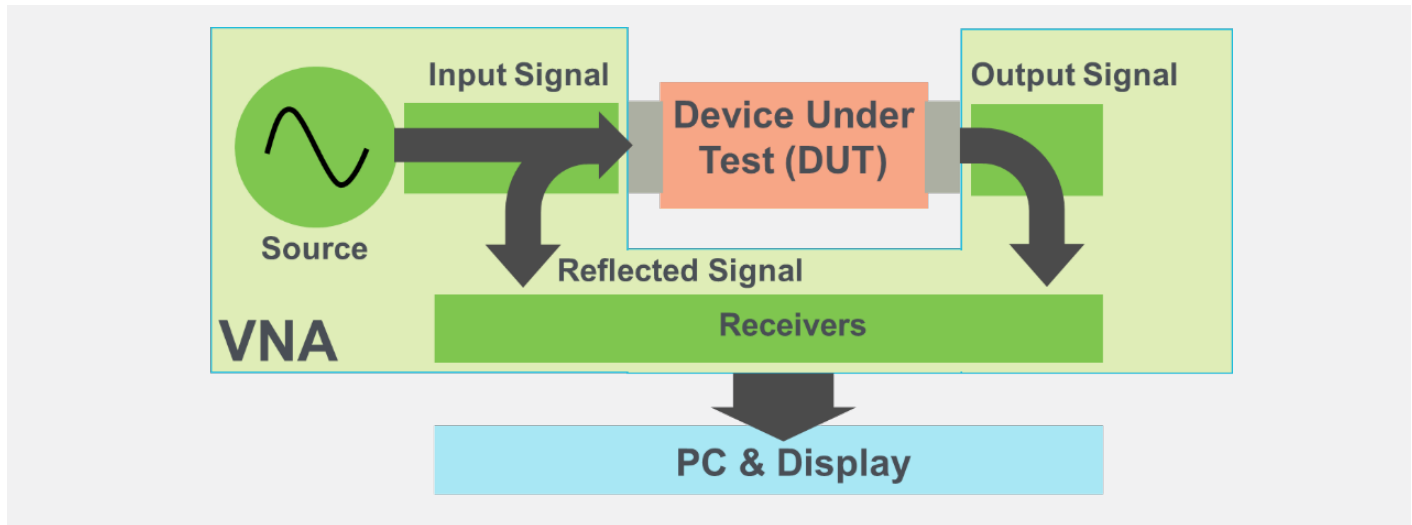


FIGURE 5. VNAs contain both a stimulus source and receivers to provide a very accurate closed loop for evaluating DUTs.

BASIC VNA OPERATION

One unique feature of a VNA is that it contains both a source, used to generate a known stimulus signal, and a set of receivers, used to determine changes to this stimulus caused by the device-under-test or DUT. Figure 5 highlights the basic operation of a VNA. For the sake of simplicity, it shows the source coming from Port 1, but most VNAs today are multipath instruments and can provide the stimulus signal to either port.

The stimulus signal is injected into the DUT and the VNA measures both the signal that's reflected from the input side, as well as the signal that passes through to the output side of the DUT. The VNA receivers measure the resulting signals and compare them to the known stimulus signal. The measured results are then processed by either an internal or external PC and sent to a display.

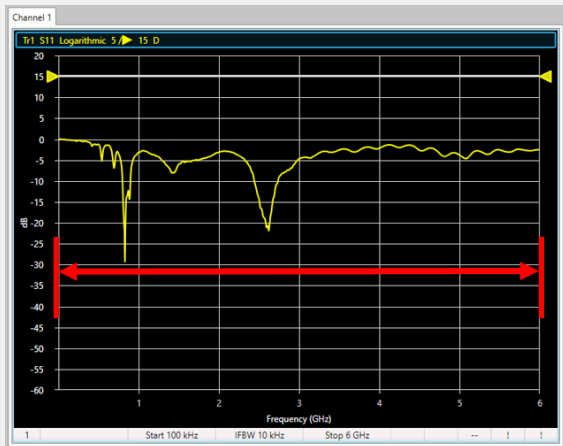
There are a variety of different VNAs available on the market, each with a different number of ports and paths for which the stimulus signal flows. In the case of a 1-port VNA, the DUT is connected to the input side of Figure 5 and only the reflected signals can be measured. For a 2-port 1-path VNA, both the reflected and transmitted signal (S_{11} and S_{21}) can be measured, however, the DUT must be physically reversed to measure the reverse parameters (S_{22} and S_{12}). As regards

to a 2-port 2-path VNA, the DUT can be connected to either port in either direction because the instrument has the capability of reversing the signal flow so that the reflections at both ports (S_{11} and S_{22}), as well as the forward and reverse transmissions (S_{21} and S_{12}), can be measured.

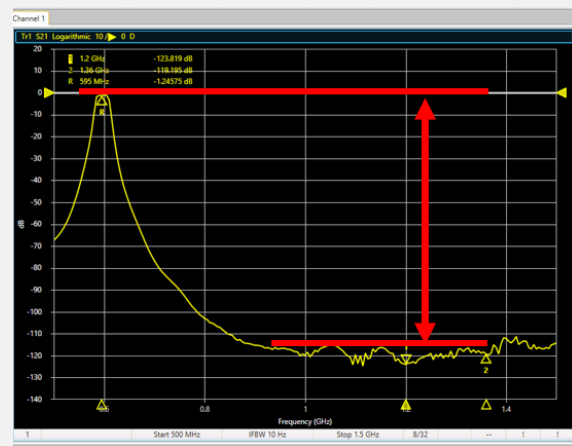
KEY SPECIFICATIONS

When determining your needs for a VNA, there are several key specifications to consider. While there are many VNA specifications, there are four top level specs which can be used to guide your selection process – frequency range, dynamic range, trace noise, and measurement speed.

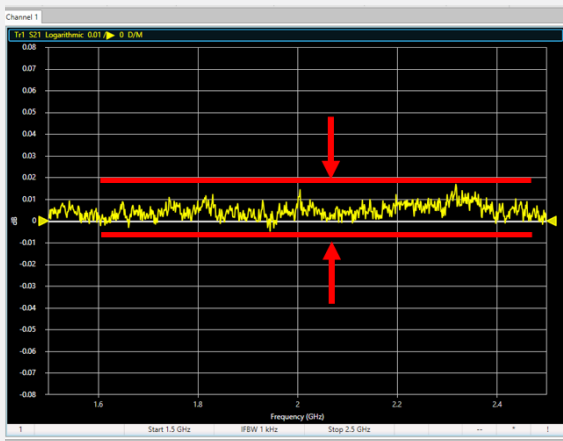
Frequency range is the first and most critical specification to consider (Figure 6a). For this, it is often good to consider not only your immediate needs but also potential future needs. In addition, while all DUTs have a given operational frequency, for some DUTs you may need to consider harmonic frequencies as well. Active components, such as amplifiers, converters and mixers may need to be tested at their harmonic frequencies which are 2 to 5 times operational frequency. Filters and duplexers may also need to be tested at harmonics of their passband. Although a higher frequency range may be desired, maximum frequency range can be a major cost driver for VNAs.



(a) Frequency Range



(b) Dynamic Range



(c) Trace Noise



(d) Measurement speed

FIGURE 6. Top level VNA specifications can be used to quickly determine the instrument class required for your application.

Dynamic range is the measurable attenuation range from max to min for a specified frequency range (Figure 6b). Based on the desired performance of your DUT, you need to make sure that the magnitude of your maximum DUT attenuation specifications are at least three to six dB less than the VNA dynamic range specification. Most VNAs today offer very good dynamic range (~ 120 dB) which is sufficient for many applications. Some very high performance components may require more expensive VNA solutions.

Trace noise measures how much random noise is generated by the VNA and passes into the measurement. It is typically measured in milli-dB (0.001 dB). Trace noise can be a key factor in determining the accuracy of certain components (Figure 6c). An example may be the acceptable level of

ripple in the passband of a filter. If you need a certain level of performance to determine accuracy of a signal through a filter, the added VNA trace noise contribution may be a factor.

Finally, one of the other specifications to consider is measurement speed (Figure 6d). Measurement speed is the time it takes to perform a single sweep or measurement. This can be the most critical requirement for high volume manufacturing applications. If you consider a component that is used in a smartphone, there may be billions of components made each year. Reducing the test time at very high volumes is critical to the success of that component. However, for many R&D and low-volume production applications, the VNA measurement speed is not an issue.

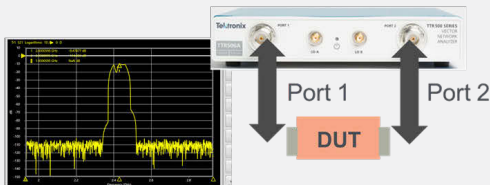
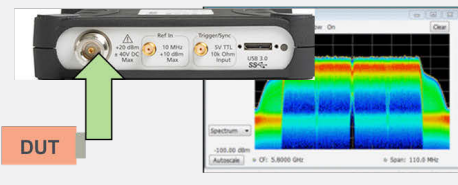
Network Analyzer	Spectrum Analyzer
 - Contains a source and receiver(s) - Uses known stimulus 2 channels or more - Ratioed measurements - Offers advanced calibration, more accurate - Limited to analog and pulsed signals	 - Contains a receiver - Measures unknown signals - Single channel - No ratioed measurements - Limited calibration capability, less accuracy - Works well with digitally modulated signals

TABLE 1. Comparing a VNA and a Spectrum Analyzer

VNA VS. SPECTRUM ANALYZER

Some design engineers may have prior experience with either a VNA or a spectrum analyzer. Others may be new to RF testing and not familiar with either. The VNA and spectrum analyzer are two of the most commonly used RF test instruments. But what's the difference between a network analyzer and a spectrum analyzer? When would you need one or both instruments? Table 1 provides a comparison of each instrument.

First, it is important to consider what type of signals you need to measure. Spectrum analyzers are the instrument of choice when measuring digitally modulated signals. If the goal is to measure, for example, the performance of Wi-Fi and LTE signals, only a spectrum analyzer can perform these measurements.

As previously mentioned, a VNA contains both source(s) and receivers. This gives it the capability to use a known stimulus to excite the DUT, and multiple receivers to measure its response. VNAs can have multiple channels and ports which allow its receivers to measure the inputs and outputs of DUTs simultaneously.

Spectrum analyzers are typically used to measure unknown signals, which may be over the air via an antenna or the output of a component. They also tend to be single channel instruments, able to measure only one output from a DUT at a

time. On the other hand, VNAs do not measure signals. They measure the inherent RF characteristics of passive or active devices.

With the known stimulus and multiple receivers, the VNA can accurately measure both the magnitude and phase characteristics of the DUT. This vector information is what allows for complete device characterization. Greater accuracy and dynamic range can also be achieved using vector error correction. This unique user calibration capability, which will be discussed later, allows VNAs to factor out the influence of cables, adaptors, and fixtures.

Some spectrum analyzers offer built-in tracking generators (SA w/TG), thus giving them much of the same capabilities as a VNA. And fundamentally speaking, a VNA works much the same way that an SA w/ TG does. However, the key difference between the two instrument solutions is the VNA's ability to measure ratioed measurements using multiple receivers. The SA w/TG does a good job for 1-port reflection measurements and can perform error correction as well. However, for transmission measurements made with the SA w/TG, measurements can be made but not with the accuracy of the VNA. Much of this, as we'll discuss later, is because full 2-port error correction is only possible on the VNA. On top of this, the majority of SA w/TGs do not display phase data, which is vital in many RF test applications.

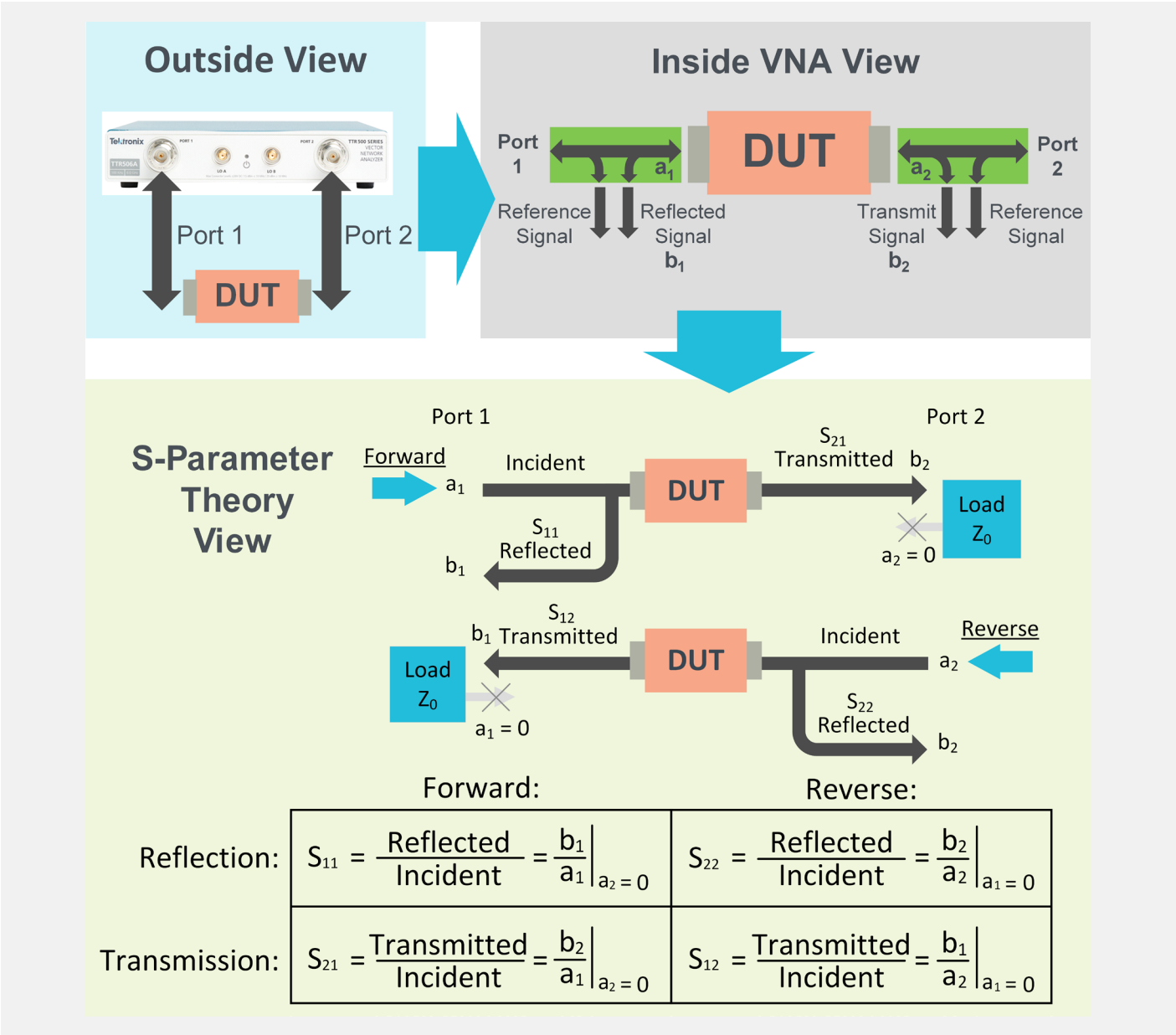


FIGURE 7. Understanding S-parameters.

Understanding S-Parameters

Since it is generally difficult to measure current or voltage at high frequencies, scattering parameters or S-parameters are measured instead. They are used to characterize the electrical properties or performance of an RF component or network of

components, and are related to familiar measurements such as gain, loss, and reflection coefficient. To understand how to use a VNA to characterize a DUT, it's important to understand the basics of S-parameters. Figure 7 walks through a simple process of explaining S-parameters.

If we start with the **Outside View**, a VNA typically has two or more ports that simply connect to the DUT - either directly or with the use of cables and adaptors. These ports are labeled, in this case, Port 1 and Port 2.

Next, let's consider the **Inside View**. The common practice used to evaluate the behavior of a multi-port network is to use incident waves as excitations at each port and to measure the resulting exiting waves that are either reflected from the port where power is applied or transmitted through the device to the remaining ports. Generally speaking, the waves entering a network or DUT are called incident waves, and the waves exiting a network or DUT are called reflected waves, although each may be composed of a combination of reflections and transmissions from other ports.

The incident waves are designated as a_n and the reflected waves are designated as b_n where n is the port number. Both a and b waves are phasors, having both magnitude and phase at the specified terminals of the network port.

Behind each of the two VNA port connectors is a directional coupler (green boxes in Figure 7). These directional couplers pass the known stimulus signal into either side of the DUT (either a_1 or a_2).

First, a portion of the stimulus signal is taken as a reference signal. S-parameters are defined as ratios of signals coming from various ports relative to this reference. At the same time, some of the stimulus signal is reflected as it enters the DUT (b_1). The portion of the input signal that is reflected is measured with a receiver connected to Port 1 inside the VNA. The portion of the input signal that enters the DUT generally experiences changes in magnitude and phase as it passes through. The portion that is emitted from port 2 is measured by the VNA receiver on Port 2 (b_2).

It's important to note that since the VNA is a bidirectional instrument, Port 2 could also be where the known stimulus is emitted (in that case a_2), and the measurement process is the same going in the reverse direction.

So now that we know more about how a VNA operates, let's **translate the Inside View into the S-parameter Theory View**. By using a (incident) and b (reflective) waves a linear network or DUT can be characterized by a set of equations describing the reflected waves from each port in terms of the incident waves at all of the ports. The constants that characterize the network under these conditions are called S-parameters.

In the *Forward* case, depicted in Figure 7, Port 1 is transmitting the a_1 signal and a matched load is applied to Port 2, resulting in zero signal reflection at the load ($a_2 = 0$). S_{11} corresponds to the reflection coefficient at Port 1, or ratio of b_1 over a_1 . S_{21} is the forward transmission coefficient through the DUT and is the ratio of b_2 over a_1 .

In the *Reverse* case, Port 2 is transmitting the a_2 signal and a matched load is applied to Port 1 ($a_1 = 0$). S_{22} corresponds to the reflection coefficient at Port 2, or ratio of b_2 over a_2 . S_{12} is the reverse transmission coefficient through the DUT and is the ratio of b_1 over a_2 .

Note that in the S-parameter nomenclature, S_{yx} , the second number (x) represents the originating port, while the first number is the destination port (y). Theoretically speaking, S-parameter theory can be applied to networks with an infinite number of ports. For example, a 4-port VNA would have 16 S-parameters: from S_{11} , S_{12} , S_{13} , S_{14} , S_{21} ... S_{44} . These S-parameters follow the same theory and are ratio measurements between each of the specified ports.

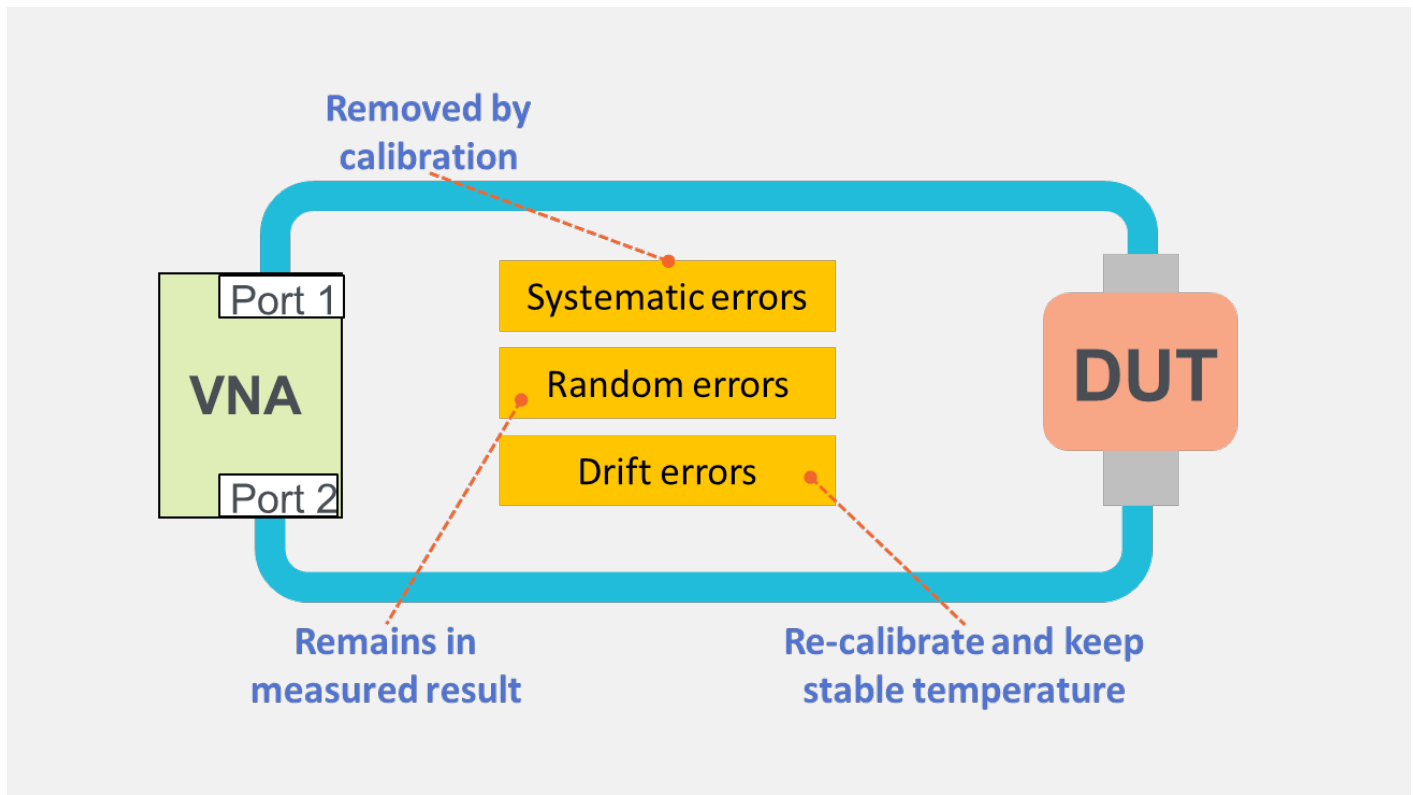


FIGURE 8. Types of VNA measurement error.

TYPES OF MEASUREMENT ERROR

Before you can make any measurements with the VNA, you must calibrate it to reduce errors that can affect the measurement. An understanding of measurement error is useful before proceeding to calibrate a VNA because not all errors can be minimized this way.

There are three main types of measurement error (Figure 8). The types of measurement error include systematic errors, random errors, and drift errors. Systematic errors are imperfections in the test equipment or in the test setup and are typically predictable. Some examples include output power variations or ripples in the VNA receiver's frequency response across its frequency range. Equally important is the power loss of RF cables that connect the DUT to the VNA that increase with frequency. Because these errors are predictable and are imperfections in the equipment, they can be easily factored out by a user calibration.

The second source of measurement error is caused by random error. This is error caused by noise emitted from the test equipment or test setup that varies with time. This error quantity is important because it will remain in the measured result even after a user calibration has been performed, and it determines the degree of accuracy that can be achieved in your measurement. Trace noise, which was discussed earlier, is an example of random error.

A third source of error is drift error, which relates to measurement drift over time. These are variances that occur in test equipment and in the test setup after a user calibration is performed. Examples are temperature fluctuations, humidity fluctuations and mechanical movement of the setup. Temperature and humidity controlled rooms are sometimes used to reduce drift error over time. The amount that the test setup drifts over time determines how often your test setup needs to be recalibrated.

Calibration Techniques

WHAT IS USER CALIBRATION

Among RF and microwave test equipment, VNAs have unique calibration techniques. While VNAs are similar to other RF and microwave test equipment in that they come factory calibrated and often require an annual check-up to be sure that they are still operating properly, VNAs are different in that they have an additional “user calibration” that can be performed by the user prior to making a measurement. Figure 9 shows the different reference planes for the factory and user calibration.

Factory calibrations cover the performance of the VNA at the test port connectors. The instrument performance is based on an input signal that meets a defined set of parameters (frequency, power, etc.) In the case of the VNA, not only is it calibrated to accurately measure from a receiver point of view, it also has a factory calibration to make sure the known stimulus from the VNA is specified and operating properly.

Basically, it ensures that the output signal meets the specs and that input signals will be represented accurately. This factory calibration is similar to the factory calibration performed on a spectrum analyzer with a tracking generator.

Having a known stimulus and receivers built within the same instrument gives the VNA a unique capability to perform an additional “user calibration”. As previously discussed, the VNA measures both magnitude and phase, which means that the user calibration performs a vector error correction. This is what makes the VNA one of the most accurate RF test instruments available. User calibration enables the VNA to factor out the effects of cables, adaptors, and most things used in the connection of the DUT. By removing the influence of the accessories, the user calibration allows for the exact measurement of the DUT performance alone. This enables designers to better understand DUT performance when it is placed into a subsystem.

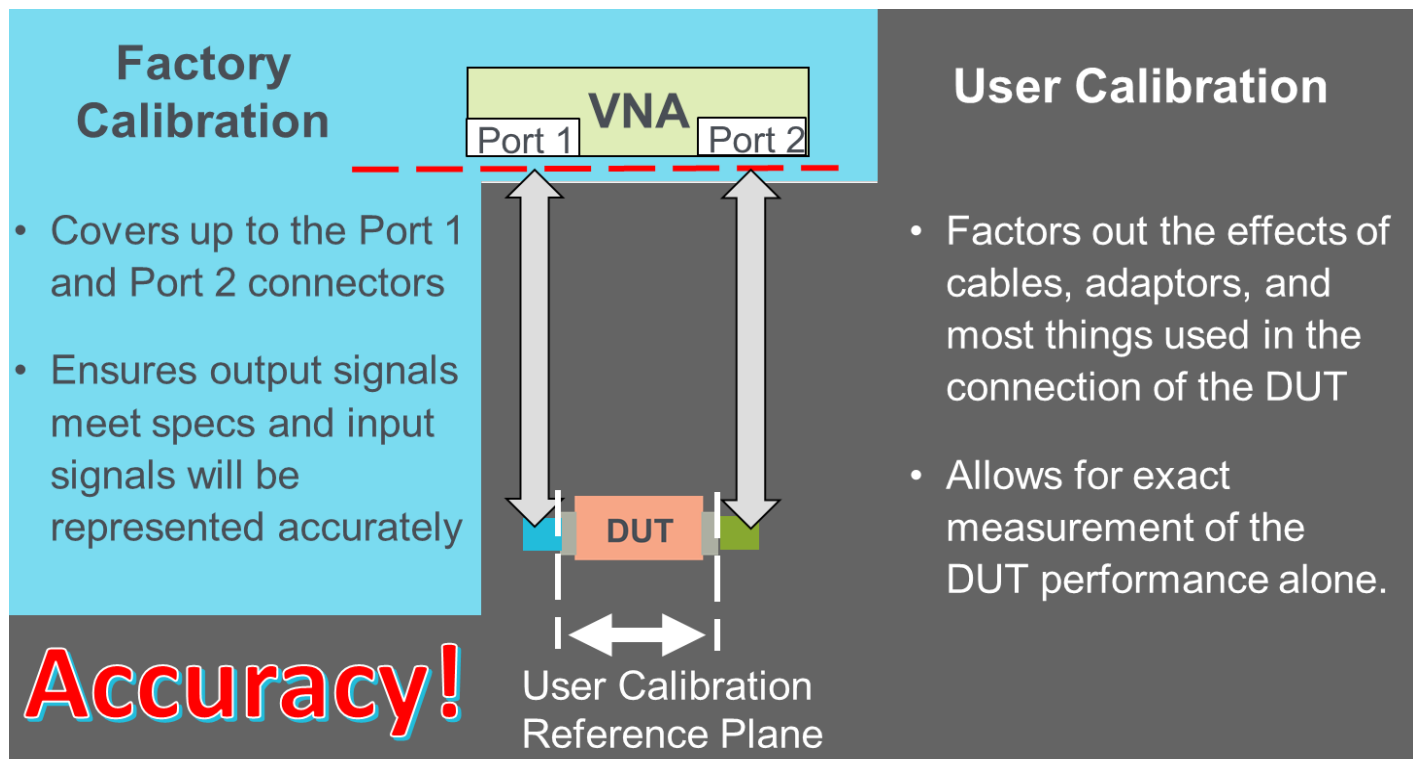


FIGURE 9. VNAs offer both factory and user calibrations.

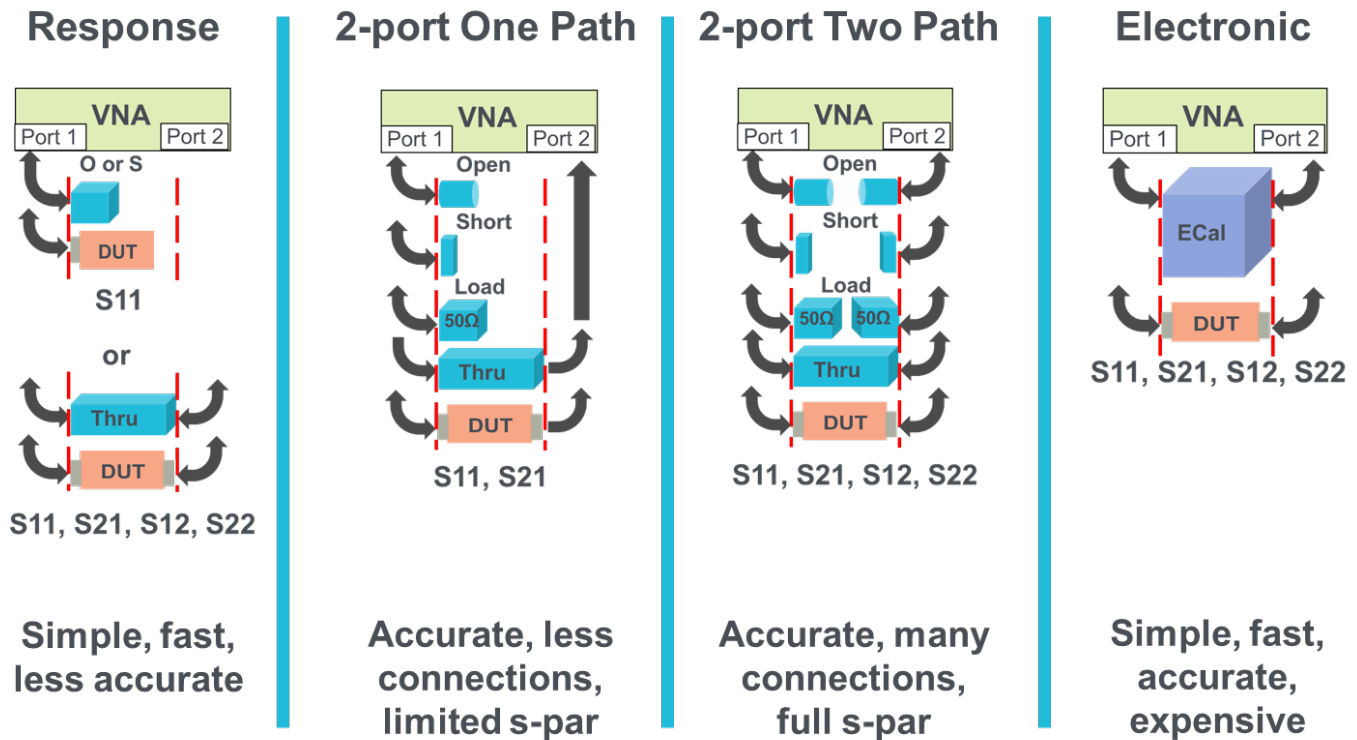


FIGURE 10. VNA calibration methods.

VNA CALIBRATION METHODS

Now that we understand the importance of the “user calibration” in factoring out measurement error, we can go ahead and discuss the different user calibration methods available. There are many different methods of VNA calibration and the complexity that you need is dependent upon your required accuracy and perhaps even your budget (Figure 10). In this section, we review some of the more common methods.

The simplest method is a response calibration. It is fast and easy, but less accurate than other methods. For example, if you only require an S_{11} or reflection measurement, you may use either an open or a short to measure the test setup response. If only an S_{21} or transmission measurement is needed, you could use only a thru standard. The response cal is easy to perform and, depending on the accuracy you need, may be sufficient.

Next, there's the 2-port one path method which is more accurate, but has fewer connections than a full 2-port two path calibration. This method works well when you're interested in a limited set of S-parameters (e.g. S_{11} , S_{21} , $a_2=0$). In this case, the VNA will only transmit from Port 1. The benefit is fewer connections during calibration.

The 2-port two path calibration method is essentially the same as the 2-port one path calibration, but with the addition of the open short load measurement on the Port 2 side. This method provides an accurate, full S-parameter measurement capability. The downside is that it requires many connections to be made. The additional steps can lead to potential process errors as you need to measure and replace standards multiple times.

Finally, there is the electronic calibration method. Simply connect the electronic calibration standard and the VNA performs a simple, fast, and very accurate calibration for S_{11} , S_{21} , S_{12} and S_{22} – all with a single set of connections. This single connection is valuable as it reduces the likelihood of inserting the wrong standard during the calibration process. Typically, an electronic calibration standard is the most expensive calibration method available. However, they add tremendous value by greatly simplifying the calibration process, while providing highly accurate results.

CALIBRATION STANDARDS

There are several types of VNA calibration standards used in the user calibration depending on the type of calibration method. The most common calibration standard set is referred to as Short, Open, Load, and Thru (SOLT). A VNA user calibration is performed using these known standards with a short circuit, open circuit, a precision load (usually 50 ohms) and a thru connection. It is best if the calibration standard has the same connector type and gender as the DUT. This allows for the DUT or calibration standard to be the only change between calibration and measurement.

Unfortunately, it is not possible to make a perfect calibration standard. A short circuit will always have some inductance; an open circuit will always have some fringing capacitance. The VNA stores data about a particular calibration kit and automatically corrects for these imperfections. The definitions of the standards for a particular calibration kit are dependent on the frequency range of the VNA. In some calibration kits, the data on the male connector is different from the female connectors, so the user may need to specify the sex of the connector within the user interface of the VNA prior to calibrating.

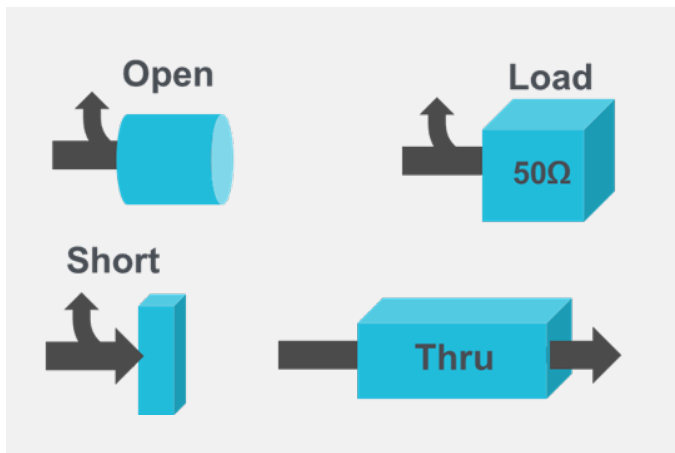


FIGURE 11. Calibration standards often include a short, open, load and thru.

The calibration standards can be physically realized in several different ways (Figure 12). Individual mechanical standards were introduced first, with each standard individually manufactured and characterized. Individual standards offer excellent accuracy and offer flexibility for a variety of test setups.



Individual Mechanical Standards

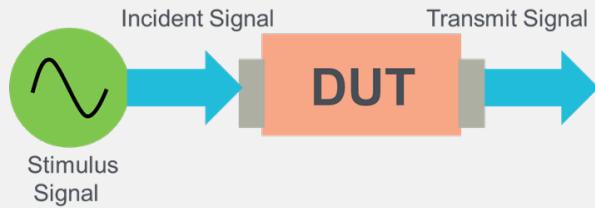


4-in-1 Mechanical Standards

FIGURE 12. Types of VNA calibration standards.

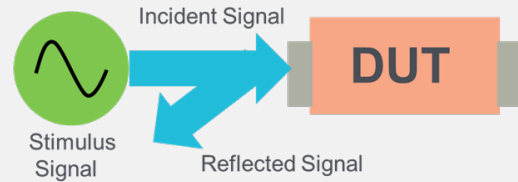
Today, 4-in-1 mechanical calibration kits are available with the open short load and thru integrated into a single mechanical device. As explained earlier, there are also automated electronic calibration standards which are driven by both a computer and a USB. These provide an automatic calibration that is very accurate and less prone to human error by reducing calibration to a single set of connections.

Transmission Measurements



- Transmission Coefficients (S_{21} , S_{12})
- Gain
- Insertion Loss/Phase
- Electrical Length/Delay
- Group Delay

Reflection Measurements



- Reflection Coefficients (S_{11} , S_{22})
- Return Loss
- VSWR (Voltage Standing Wave Ratio)
- Impedance ($R+jX$)

FIGURE 13. VNAs perform transmission and reflection measurements.

Typical VNA Measurements

VNA's perform two types of measurements – transmission and reflection (Figure 13). Transmission measurements pass the VNA's stimulus signal through the DUT, which is then measured by the VNA receivers on the other side. The most common transmission S-parameter measurements are S_{21} and S_{12} (S_{xy} for greater than 2-ports). Swept power measurements are a form of transmission measurement. Some other examples of transmission measurements include gain, insertion loss/phase, electrical length/delay and group delay. Comparatively, reflection measurements measure the part of the VNA stimulus signal that is incident upon the DUT, but does not pass through it. Instead, the reflection measurement measures the signal that travels back towards the source due to reflections. The most common reflection S-parameter measurements are S_{11} and S_{22} (S_{xx} for greater than 2-ports).

SWEPT FREQUENCY MEASUREMENTS

Swept frequency measurements are particularly useful because they sweep the internal source across a user defined set of frequencies and step points. A wide variety of measurements can be made from this including S-parameters, individual incident and reflected waves (e.g. a_1 , b_2), magnitude, phase, etc. Figure 14 shows an example of a swept frequency

transmission measurement of a passive filter. This type of filter measurement shows what happens to the signal as it passes through the component. The S_{21} measurement indicates the passband bandwidth performance as defined by its 6 dB response. The stopband performance is displayed as compared to a 60 dB reduction specification. The measured result can then be compared with the filter design goals or, from the system designer's perspective, the filter manufacturer's specification.

Example: Passive Filter

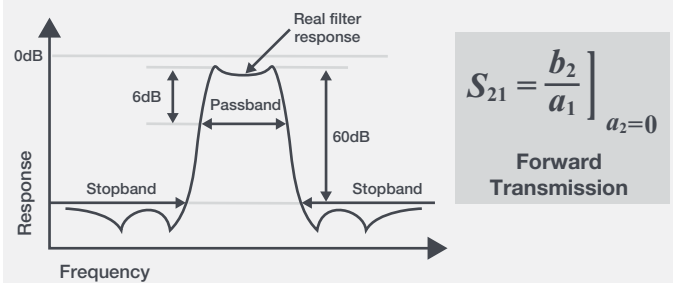


FIGURE 14. Swept frequency transmission measurement example of a passive filter.

Swept frequency measurements may also measure reflections of the stimulus signal that are incident on the DUT, but are reflected as opposed to being transmitted through the DUT. These S_{11} (or S_{xx}) measurements allow the user to check and compare the performance of the DUT to its specification. Example DUTs include antennas, filters, and duplexers. Figure 15 shows an example of an antenna return loss measurement. Note that in the antenna passband, most of the signal is being transmitted so a visible null occurs in the reflection measurement result.

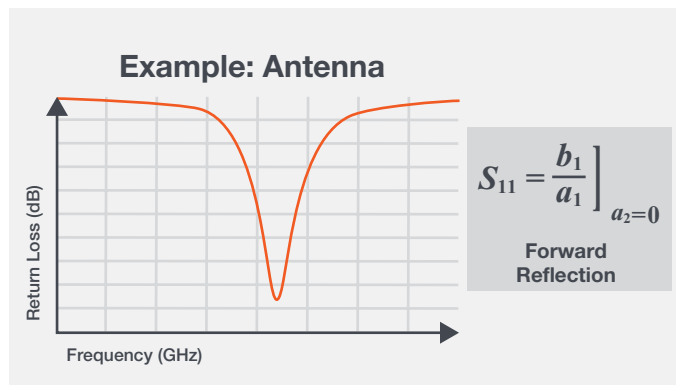


FIGURE 15. Swept frequency reflection measurement example of an antenna.

TIME DOMAIN MEASUREMENTS

Some VNAs are capable of using inverse Fourier transforms to convert swept frequency measurements into the time domain. In this way, data displayed in the time domain allows the VNA to be used to find problems in cables and connections by detecting the locations of impedance mismatches or discontinuities as the signal passes through the DUT.

For time domain measurements, the ability to resolve two signals is inversely proportional to the measured frequency span. Therefore, the wider the frequency span, the greater the ability the VNA has to distinguish between closely spaced discontinuities. The maximum frequency span is set by the user and may be defined by either the frequency range of the VNA or the viable bandwidth of the DUT.

The data collected in the frequency domain is not continuous, but a finite number of discrete frequency points. This causes the time domain data to repeat after the inverse of the frequency sample interval. This phenomenon is called aliasing. It is important to set the frequency sample interval correctly to measure the required distance accurately to evaluate the DUT's performance before aliasing occurs.

Figure 16 shows a VNA measurement of a cable with several adapters. This could be a base station cable running from the base station subsystem to its antenna. The time domain measurement locates the physical distance to the different adapters or potential discontinuities in the cable, which helps locate problem areas or faults.

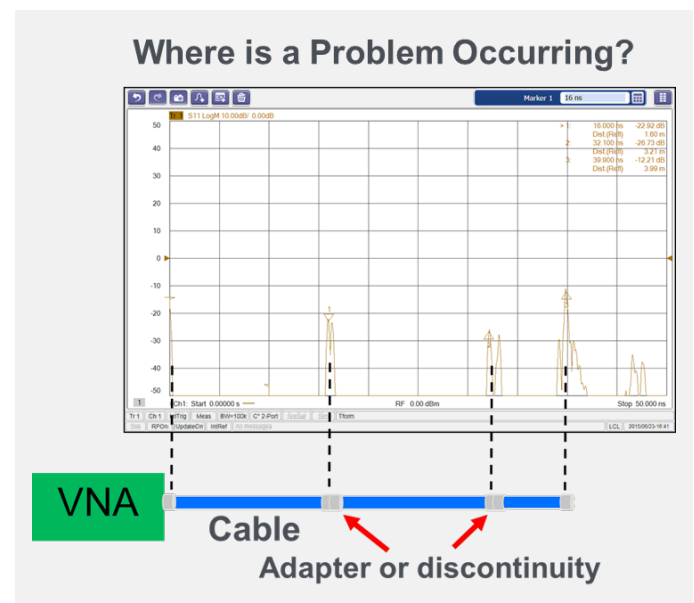


FIGURE 16. VNAs mathematically convert swept frequency measurements into the time domain. The measurements can be useful for locating impedance mismatches or faults in the line.

SWEPT POWER MEASUREMENTS

Instead of sweeping frequencies, VNAs may also sweep the stimulus signal's output power level. For these measurements, the frequency is held constant while the output power is incrementally stepped across a defined power range. This is a common measurement for amplifiers, starting at a low power level and incrementing the power at fractional dB steps.

In the linear region of an amplifier, as the input power increases, the output power increases proportionately. The point when the amplifier output deviates from the linear expectation by 1 dB is referred to as the 1 dB compression point (Figure 17). When the amplifier reaches its compression point, it is no longer able to increase its output power as before. For applications that require linear performance of an amplifier, this measurement helps define that specification.

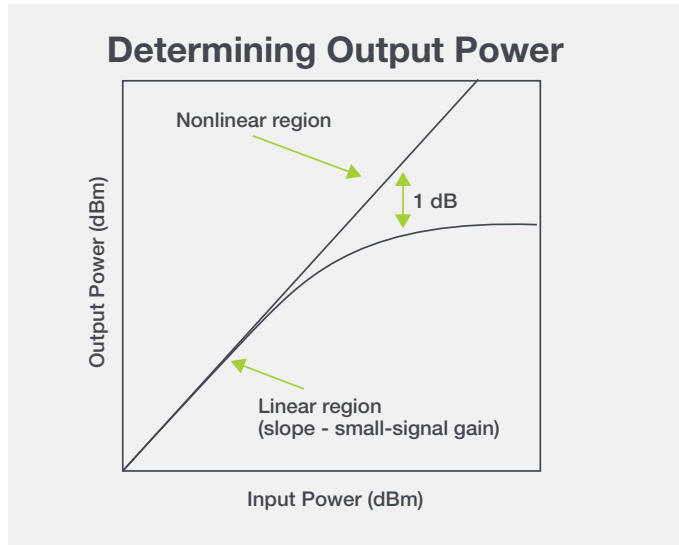


FIGURE 17. Swept power measurements are commonly done on amplifiers.

TESTING MULTIPORT COMPONENTS

Many components today have more than two ports (Figure 18). They may have one input and multiple outputs or vice versa. More complex components can have multiple inputs and multiple outputs. If the interaction between the ports is not a concern, some of these components may still be tested with a series of 2-port measurements.

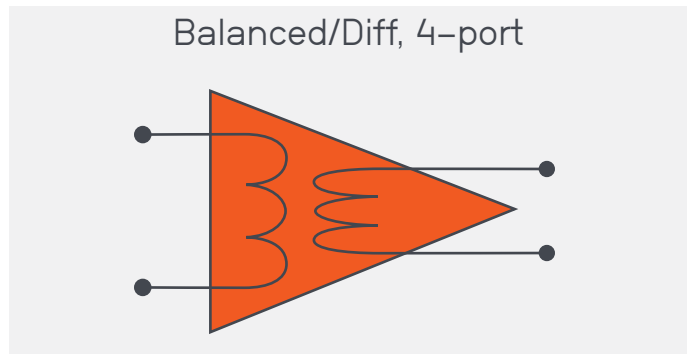


FIGURE 18. Many components today have more than 2-ports.

When there's a need to measure the interaction between multiple ports, you may need a multiport VNA. A true multiport measurement would measure N^2 S-parameters and require a VNA with N-ports, where N equals the number of DUT ports. Instead of only S_{11} , S_{21} , S_{12} , and S_{22} , the S-parameters would also include S_{41} or S_{43} or $S_{10\ 11}$, for example. The true multiport VNA can provide a stimulus signal to each of the ports. Multiport error correction removes systematic errors for the measurement, but requires a complex calibration process where calibration standards must be connected to all possible combinations of ports.

Summary

Now, it is easy to understand why VNAs have helped to make many modern technologies possible. By providing a known stimulus signal to the device under test or DUT, and multiple receivers to measure the response, the VNA forms a closed loop, allowing it to measure the electrical magnitude and phase response of components very accurately. And due to its unique user calibration, the VNA is one of the most accurate RF test instruments available. It allows for careful isolation of the DUT performance by reducing the influence of cables, adapters and other testing aides.

VNAs test component specifications and verify design simulations. With this accurate level of characterization, system engineers can study a circuit or system-level design and rest assured knowing – from the design phase to manufacturing phase – it's going to function as expected.

